High Performance Low-Power, LDO Regulator with Enable

The NCV8560 provides 150 mA of output current at fixed voltage options, or an adjustable output voltage from 5.0 V down to 1.250 V. It is designed for portable battery powered applications and offers high performance features such as low power operation, fast enable response time, and low dropout.

The device is designed to be used with low cost ceramic capacitors and is packaged in the DFN6, 3x3 and TSOP-5 packages.

Features

- Output Voltage Options:
 Adjustable, 1.3 V, 1.5 V, 1.8 V, 2.5 V, 2.8 V, 3.0 V, 3.3 V, 3.5 V, 5.0 V
- Ultra-Low Dropout Voltage of 150 mV at 150 mA
- Adjustable Output by External Resistors from 5.0 V down to 1.250 V
- Fast Enable Turn-on Time of 15 μs
- Wide Supply Voltage Range Operating Range
- Excellent Line and Load Regulation
- High Accuracy up to 1.5% Output Voltage Tolerance over All Operating Conditions
- Typical Noise Voltage of 50 μV_{rms} without a Bypass Capacitor
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable
- These Devices are Pb-Free and are RoHS Compliant

Typical Applications

- SMPS Post-Regulation
- Hand-held Instrumentation
- Noise Sensitive Circuits VCO, RF Stages, etc.
- Camcorders and Cameras

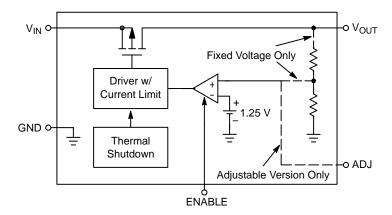


Figure 1. Simplified Block Diagram



ON Semiconductor®

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MN SUFFIX CASE 488AE DFN6 MN SUFFIX CASE 506BA TSOP-5 SN SUFFIX CASE 483

MARKING DIAGRAMS

DFN6



V8560 or AQ = Specific Device Code xxx = ADJ, 150, 180, 250, 280,

= ADJ, 150, 180, 250, 28 300, 330, 350 or 500

A = Assembly Location

L = Wafer Lot Y = Year W = Work Week

M = Date CodePb-Free Package

TSOP-5



xxx = Specific Device Code

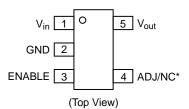
A = Assembly Location
Y = Year
W = Work Week
= Pb-Free Package

(Note: Microdot may be in either location)

ORDERING INFORMATION

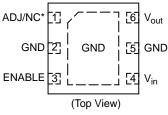
See detailed ordering and shipping information in the package dimensions section on page 13 of this data sheet.

PIN CONNECTIONS



* ADJ – Adjustable Version NC – Fixed Voltage Version

Figure 2. TSOP5



* ADJ – Adjustable Version NC – Fixed Voltage Version

Figure 3. DFN6 (3x3)

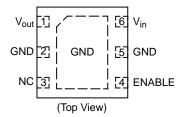


Figure 4. DFN6 (2x2.2)

PIN FUNCTION DESCRIPTION

	Pin No.			
DFN6 (2x2.2)	DFN6 (3x3)	TSOP-5	Pin Name	Description
3	1	4	ADJ/NC	Output Voltage Adjust Input (Adjustable Version), No Connection (Fixed Voltage Versions) (Note 1)
2, 5, EPAD	2, 5, EPAD	2	GND	Power Supply Ground; Device Substrate
4	3	3	ENABLE	The Enable Input places the device into low–power standby when pulled to logic low (< 0.4 V). Connect to $V_{\rm in}$ if the function is not used.
6	4	1	V _{in}	Positive Power Supply Input
1	6	5	V _{out}	Regulated Output Voltage

^{1.} True no connect. Printed circuit board traces are allowable.

ABSOLUTE MAXIMUM RATINGS

Rating	Symbol	Min	Max	Unit
Input Voltage (Note 2)	V _{in}	-0.3	6	V
Output, Enable, Adjustable Voltage	V _{out} , ENABLE, ADJ	-0.3	V _{in} + 0.3 V	V
Maximum Junction Temperature	T _{J(max)}	-	150	°C
Storage Temperature	T _{STG}	-65	150	°C
ESD Capability, Human Body Model (Note 3)	ESD _{HBM}	3500	_	V
ESD Capability, Machine Model (Note 3)	ESD _{MM}	200	_	V
Moisture Sensitivity Level	MSL	MSL1/260		_

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

- 2. Refer to ELECTRICAL CHĂRACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.
- 3. This device series incorporates ESD protection and is tested by the following methods:
 - ESD Human Body Model tested per AEC-Q100-002 (EIA/JESD22-A114)
 - ESD Machine Model tested per AEC-Q100-003 (EIA/JESD22-A115)
 - Latchup Current Maximum Rating: ≤150 mA per JEDEC standard: JESD78.

THERMAL CHARACTERISTICS

Rating	Symbol	Value	Unit
Thermal Characteristics, DFN6, 3x3.3 mm (Note 4) Thermal Resistance, Junction–to–Air (Note 5)	$R_{ hetaJA}$	107	°C/W
Thermal Characteristics, TSOP–5 (Note 4) Thermal Resistance, Junction–to–Air (Note 5)	$R_{ hetaJA}$	205	°C/W
Thermal Characteristics, DFN6, 2x2.2 mm (Note 4) Thermal Resistance, Junction–to–Air (Note 5)	$R_{ hetaJA}$	122	°C/W

- 4. Refer to ELECTRICAL CHARACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.
- 5. As measured using a copper heat spreading area of $650\ mm^2$, 1 oz copper thickness.

OPERATING RANGES

Rating	Symbol	Min	Max	Unit
Operating Input Voltage (Note 6)	V _{in}	V _{out} + V _{DO} , 1.75 V (Note 7)	6	V
Adjustable Output Voltage Range (Adjustable Version Only)	V _{out}	1.25	5.0	V
Operating Ambient Temperature Range	T _A	-40	125	°C

Refer to ELECTRICAL CHARACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.
 Minimum V_{in} = 1.75 V or (V_{out} + V_{DO}), whichever is higher.

ELECTRICAL CHARACTERISTICS (V_{in} = 1.750 V, V_{out} = 1.250 V (adjustable version)), (V_{in} = V_{out} + 0.5 V (fixed version)), C_{in} = C_{out} =1.0 μ F, -40° C \leq T_{A} \leq 125 $^{\circ}$ C, Figure 5, unless otherwise specified.) (Note 8)

Characteristic	Symbol	Test Conditions	Min	Тур	Max	Unit
GENERAL	•		•		•	•
Disable Current	I _{DIS}	ENABLE = 0 V, Vin = 6 V $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$	_	0.01	1.0	μΑ
Ground Current Adjustable Option 1.3 V Option 1.5 V Option 1.8 V to 3.0 V Option 3.3 V to 5.0 V Option	I _{GND}	ENABLE = 0.9 V, I _{out} = 1.0 mA to 150 mA	- - - -	100 135 135 140 145	135 150 170 175 180	μΑ
Thermal Shutdown Temperature (Note 9)	T _{SD}		150	175	200	°C
Thermal Shutdown Hysteresis	T _{SH}		_	10	_	°C
ADJ Input Bias Current	I _{ADJ}		-0.75	-	0.75	μΑ
CHIP ENABLE						
ENABLE Input Threshold Voltage	V _{th(EN)}					V
Voltage Increasing, Logic High			0.9	-	-	
Voltage Decreasing, Logic Low			_	-	0.4	
Enable Input Bias Current (Note 9)	I _{EN}		-	3.0	100	nA
TIMING						
Output Turn On Time Adjustable Option 1.3 V to 3.5 V Option 5.0 V Option	t _{EN}	ENABLE = 0 V to V _{in}	- - -	15 15 30	25 25 50	μs

^{8.} Performance guaranteed over the indicated operating temperature range by design and/or characterization, production tested at T_J = T_A = 25°C. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.

^{9.} Values based on design and/or characterization.

ELECTRICAL CHARACTERISTICS

 $(V_{in} = 1.750 \text{ V}, V_{out} = 1.250 \text{ V}, C_{in} = C_{out} = 1.0 \ \mu\text{F}, -40^{\circ}\text{C} \le T_A \le 125^{\circ}\text{C}, \text{ Figure 5, unless otherwise specified.)}$ (Note 10)

Characteristic	Symbol	Test Conditions	Min	Тур	Max	Unit	
REGULATOR OUTPUT (Adjustable Voltage Version)							
Output Voltage	V _{out}	I _{out} = 1.0 mA to 150 mA V _{in} = 1.75 V to 6.0 V, V _{out} = ADJ	1.231 (-1.5%)	1.250	1.269 (+1.5%)	V	
Ripple Rejection $(V_{in} = V_{out} + 1.0 \text{ V} + 0.5 \text{ V}_{p-p})$	RR	I _{out} = 1.0 mA to 150 mA f = 120 Hz f = 1.0 kHz f = 10 kHz	- - -	62 55 38	- - -	dB	
Line Regulation	Reg _{line}	V _{in} = 1.750 V to 6.0 V, I _{out} = 1.0 mA	_	1.0	10	mV	
Load Regulation	Reg _{load}	I _{out} = 1.0 mA to 150 mA	-	2.0	15	mV	
Output Noise Voltage (Note 11)	V _n	f = 10 Hz to 100 kHz	_	50	_	μV_{rms}	
Output Short Circuit Current	I _{sc}		300	550	800	mA	
Dropout Voltage $ \begin{array}{c} V_{out} = 1.25 \text{ V} \\ V_{out} = 1.3 \text{ V} \\ V_{out} = 1.5 \text{ V} \\ V_{out} = 1.5 \text{ V} \\ V_{out} = 1.8 \text{ V} \\ V_{out} = 2.5 \text{ V} \\ V_{out} \ge 2.8 \text{ V} \end{array} $	V _{DO}	Measured at: V _{Out} – 2.0%, I _{Out} = 150 mA, Figure 6	- - - - -	175 175 150 125 100 75	250 250 225 175 150 125	mV	

REGULATOR OUTPUT (Fixed Voltage Version) ($V_{in} = V_{out} + 0.5 \text{ V}$, $C_{in} = C_{out} = 1.0 \mu\text{F}$, $-40^{\circ}\text{C} \le T_{A} \le 125^{\circ}\text{C}$, Figure 7, unless otherwise specified.) (Note 10)

Output Voltage 1.3 V Option 1.5 V Option 1.8 V Option 2.5 V Option 2.8 V Option 3.0 V Option 3.3 V Option 3.5 V Option 5.0 V Option	Vout	l _{out} = 1.0 mA to 150 mA V _{in} = (V _{out} + 0.5 V) to 6.0 V	1.274 1.470 1.764 2.450 2.744 2.940 3.234 3.430 4.900 (-2%)		1.326 1.530 1.836 2.550 2.856 3.060 3.366 3.570 5.100 (+2%)	V
Power Supply Ripple Rejection (Note 11) (Vin = V _{out} + 1.0 V + 0.5 V _{p-p})	PSRR	I _{out} = 1.0 mA to 150 mA f = 120 Hz f = 1.0 kHz f = 10 kHz	- - -	62 55 38	- - -	dB
Line Regulation	Reg _{line}	V _{in} = 1.750 V to 6.0 V, I _{out} = 1.0 mA	-	1.0	10	mV
Load Regulation 1.3 V to 1.5 V Option 1.8 V Option 2.5 V to 5.0 V Option	Reg _{load}	I _{out} = 1.0 mA to 150 mA	- - -	2.0 2.0 2.0	20 25 30	mV
Output Noise Voltage (Note 11)	V _n	f = 10 Hz to 100 kHz	-	50	_	μV_{rms}
Output Short Circuit Current	I _{sc}		300	550	800	mA
Dropout Voltage 1.3 V Option 1.5 V Option 1.8 V Option 2.5 V Option 2.8 V to 5.0 V Option	V _{DO}	Measured at: V _{out} – 2.0%	- - - - -	175 150 125 100 75	250 225 175 150 125	mV

^{10.} Performance guaranteed over the indicated operating temperature range by design and/or characterization, production tested at $T_J = T_A = 25^{\circ}C$. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible. 11. Values based on design and/or characterization.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

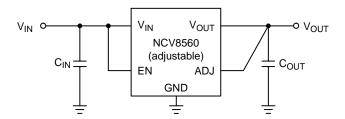


Figure 5. Typical Application Circuit for V_{out} = 1.25 V (Adjustable Version)

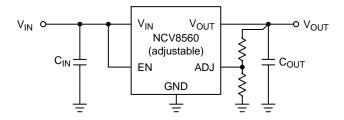


Figure 6. Typical Application Circuit for Adjustable Vout

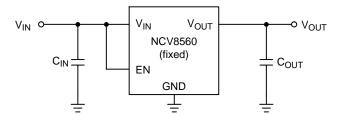
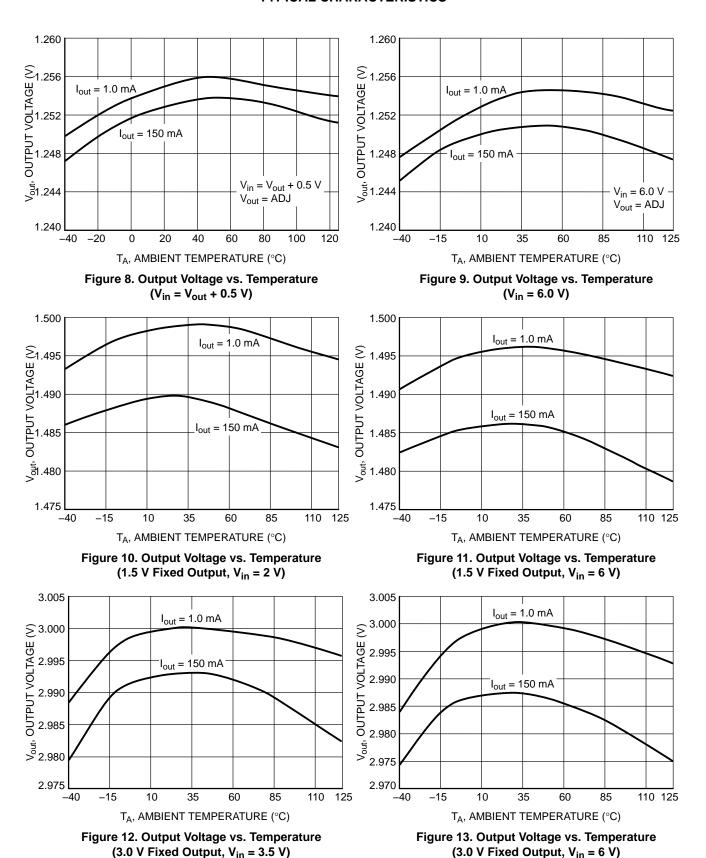
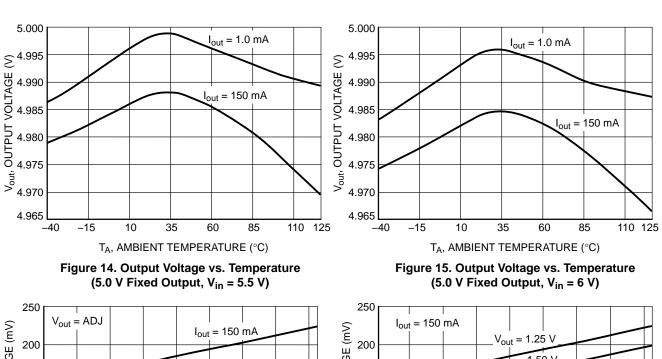


Figure 7. Typical Application Circuit (Fixed Voltage Version)





V_{DO}, DROPOUT VOLTAGE (mV) 150 100 $I_{out} = 50 \text{ mA}$ 50 $I_{out} = 1.0 \text{ mA}$ 40 60 80 -40 -20 0 20 100 120 TA, AMBIENT TEMPERATURE (°C)

Figure 16. Dropout Voltage vs. Temperature (Over Current Range)

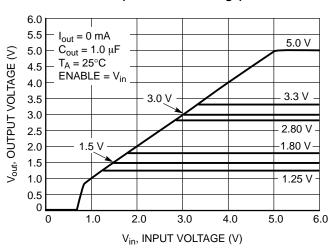


Figure 18. Output Voltage vs. Input Voltage

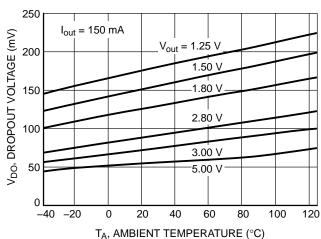


Figure 17. Dropout Voltage vs. Temperature (Over Output Voltage)

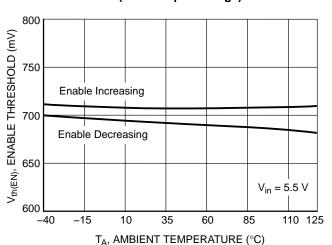


Figure 19. Enable Threshold vs. Temperature

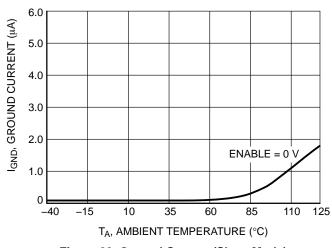


Figure 20. Ground Current (Sleep Mode) vs.
Temperature

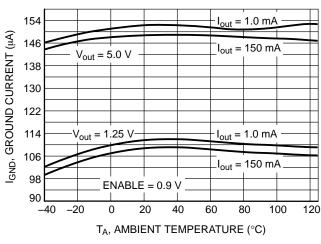


Figure 21. Ground Current (Run Mode) vs.
Temperature

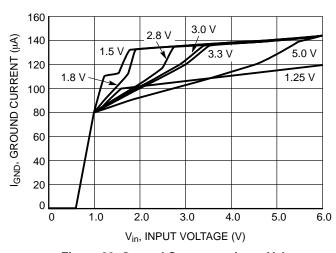


Figure 22. Ground Current vs. Input Voltage

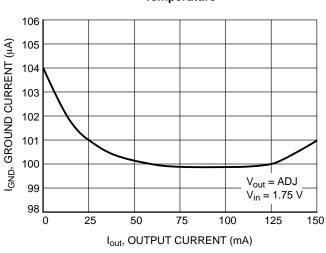


Figure 23. Ground Current vs. Output Current

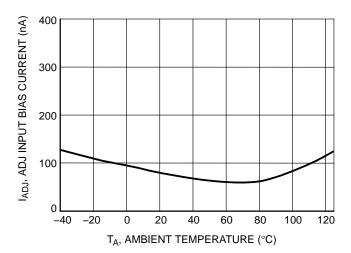


Figure 24. ADJ Input Bias Current vs. Temperature

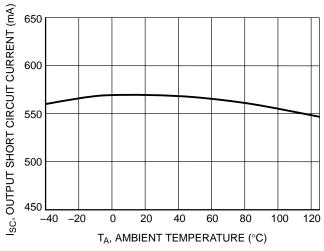


Figure 25. Output Short Circuit Current vs.
Temperature

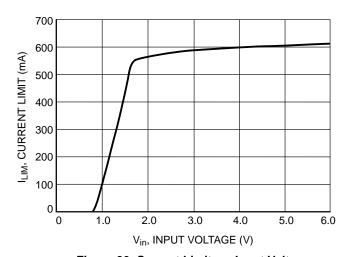


Figure 26. Current Limit vs. Input Voltage

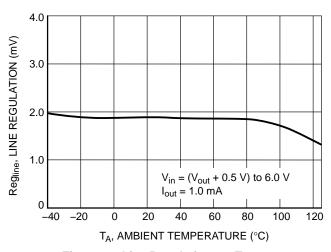


Figure 27. Line Regulation vs. Temperature

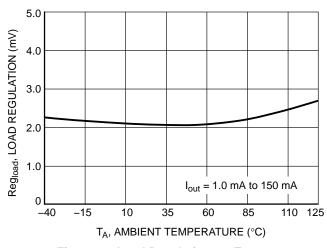


Figure 28. Load Regulation vs. Temperature

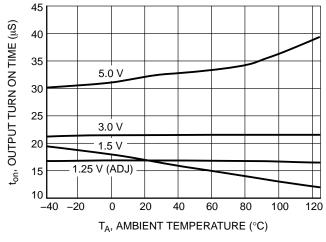


Figure 29. Output Turn On Time vs. Temperature

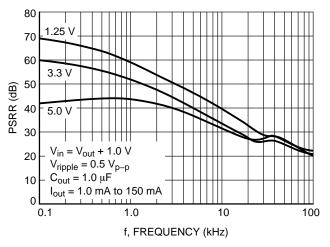


Figure 30. Power Supply Ripple Rejection vs. Frequency

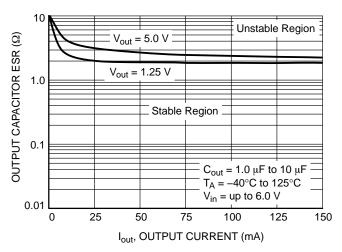


Figure 31. Output Stability with Output Capacitor ESR over Output Current

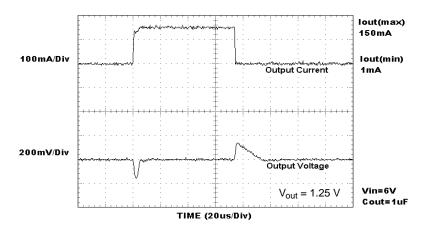


Figure 32. Load Transient Response (1.0 μF)

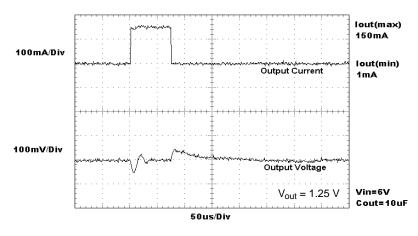


Figure 33. Load Transient Response (10 μF)

DEFINITIONS

Load Regulation

The change in output voltage for a change in output load current at a constant temperature.

Dropout Voltage

The input/output differential at which the regulator output no longer maintains regulation against further reductions in input voltage. Measured when the output drops 2% below its nominal. The junction temperature, load current, and minimum input supply requirements affect the dropout level.

Output Noise Voltage

This is the integrated value of the output noise over a specified frequency range. Input voltage and output load current are kept constant during the measurement. Results are expressed in μV_{rms} or nV/\sqrt{Hz} .

Ground Current

Ground Current (I_{GND}) is the current that flows through the ground pin when the regulator operates with a load on its output. This consists of internal IC operation, bias, etc. It is actually the difference between the input current (measured through the LDO input pin) and the output load current. If the regulator has an input pin that reduces its internal bias and shuts off the output (enable/disable function), this term is called the disable current (I_{DIS}).

Line Regulation

The change in output voltage for a change in input voltage. The measurement is made under conditions of low dissipation or by using pulse techniques such that the average junction temperature is not significantly affected.

Line Transient Response

Typical output voltage overshoot and undershoot response when the input voltage is excited with a given slope.

Load Transient Response

Typical output voltage overshoot and undershoot response when the output current is excited with a given slope between no-load and full-load conditions.

Thermal Protection

Internal thermal shutdown circuitry is provided to protect the integrated circuit in the event that the maximum junction temperature is exceeded. When activated at typically 175°C, the regulator turns off. This feature is provided to prevent failures from accidental overheating.

Maximum Package Power Dissipation

The power dissipation level at which the junction temperature reaches its maximum operating value.

APPLICATIONS INFORMATION

The NCV8560 series regulator is self-protected with internal thermal shutdown and internal current limit. Typical application circuits are shown in Figures 5 and 6.

Input Decoupling (Cin)

A ceramic or tantalum $1.0~\mu F$ capacitor is recommended and should be connected close to the NCV8560 package. Higher capacitance and lower ESR will improve the overall line transient response.

Output Decoupling (Cout)

The NCV8560 is a stable component and does not require a minimum Equivalent Series Resistance (ESR) for the output capacitor. The minimum output decoupling value is $1.0\,\mu\text{F}$ and can be augmented to fulfill stringent load transient requirements. The regulator works with ceramic chip capacitors as well as tantalum devices. Larger values improve noise rejection and load regulation transient response. Figure 31 shows the stability region for a range of operating conditions and ESR values.

No-Load Regulation Considerations

The NCV8560 adjustable regulator will operate properly under conditions where the only load current is through the resistor divider that sets the output voltage. However, in the case where the NCV8560 is configured to provide a 1.250 V

output, there is no resistor divider. If the part is enabled under no-load conditions, leakage current through the pass transistor at junction temperatures above 85°C can approach several microamps, especially as junction temperature approaches 150°C. If this leakage current is not directed into a load, the output voltage will rise up to a level approximately 20 mV above nominal.

The NCV8560 contains an overshoot clamp circuit to improve transient response during a load current step release. When output voltage exceeds the nominal by approximately 20 mV, this circuit becomes active and clamps the output from further voltage increase. Tying the ENABLE pin to V_{in} will ensure that the part is active whenever the supply voltage is present, thus guaranteeing that the clamp circuit is active whenever leakage current is present.

When the NCV8560 adjustable regulator is disabled, the overshoot clamp circuit becomes inactive and the pass transistor leakage will charge any capacitance on V_{out} . If no load is present, the output can charge up to within a few millivolts of V_{in} . In most applications, the load will present some impedance to V_{out} such that the output voltage will be inherently clamped at a safe level. A minimum load of $10~\mu A$ is recommended.

Noise Decoupling

The NCV8560 is a low noise regulator and needs no external noise reduction capacitor. Unlike other low noise regulators which require an external capacitor and have slow startup times, the NCV8560 operates without a noise reduction capacitor, has a typical 15 µs start up delay and achieves a 50 µV_{rms} overall noise level between 10 Hz and 100 kHz.

Enable Operation

The enable pin will turn the regulator on or off. The threshold limits are covered in the electrical characteristics table in this data sheet. The turn-on/turn-off transient voltage being supplied to the enable pin should exceed a slew rate of 10 mV/µs to ensure correct operation. If the enable function is not to be used then the pin should be connected to V_{in}.

Output Voltage Adjust

The output voltage can be adjusted from 1 times (Figure 5) to 4 times (Figure 6) the typical 1.250 V regulation voltage via the use of resistors between the output and the ADJ input. The output voltage and resistors are chosen using Equation 1 and Equation 2.

$$V_{out} = 1.250 \left(1 + \frac{R_1}{R_2} \right) + \left(I_{ADJ} \times R_1 \right)$$
 (eq. 1)

$$R_2 \cong \frac{R_1}{\frac{V_{out}}{1.25} - 1}$$
 (eq. 2)

Input bias current I_{ADJ} is typically less than 150 nA. Choose R2 arbitrarily to minimize errors due to the bias current and to minimize noise contribution to the output voltage. Use Equation 2 to find the required value for R1.

Thermal

As power in the NCV8560 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature rise for the part. When the NCV8560 has good thermal conductivity through the PCB, the junction temperature will be relatively low with high power applications. The maximum dissipation the NCV8560 can handle is given by:

$$P_{D(MAX)} = \frac{T_{J(MAX)} - T_{A}}{R_{\theta JA}}$$
 (eq. 3)

Since T_J is not recommended to exceed 125°C (T_{J(MAX)}), then the NCV8560 in a DFN6 package can dissipate up to 600 mW when the ambient temperature (T_A) is 25°C, and PCB area is 150mm² and larger, see Figure 34.

The power dissipated by the NCV8560 can be calculated from the following equations:

$$P_D \approx V_{in}(I_{GND} @ I_{out}) + I_{out}(V_{in} - V_{out})$$
 (eq. 4)

or

$$V_{in(MAX)} \approx \frac{P_{D(MAX)} + (V_{out} \times I_{out})}{I_{out} + I_{GND}}$$
 (eq. 5)

If a 150 mA output current is needed, the quiescent current I_{GND} is taken from the data sheet electrical characteristics table or extracted from Figure 21 and Figure 23. I_{GND} is approximately $108 \mu A$ when $I_{out} = 150 \text{ mA}$. For an output voltage of 1.250 V, the maximum input voltage will then be 3.9 V, good for a 1 Cell Li-ion battery.

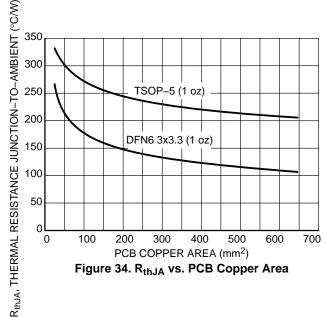


Figure 34. R_{thJA} vs. PCB Copper Area

Hints

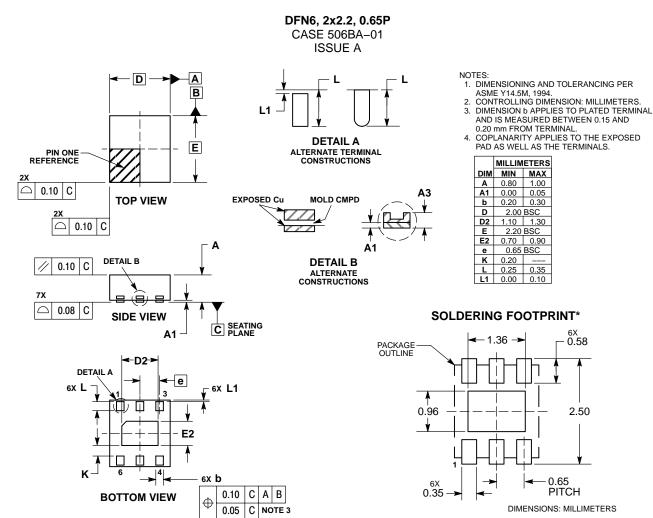
V_{in} and GND printed circuit board traces should be as wide as possible. When the impedance of these traces is high, there is a chance to pick up noise or cause the regulator to malfunction. Place external components, especially the output capacitor, as close as possible to the NCV8560, and make traces as short as possible.

DEVICE ORDERING INFORMATION

Device*	Marking Code	Version	Package	Shipping [†]
NCV8560MNADJR2G	1st Line: V8560 2nd Line: ADJ	ADJ		
NCV8560MN150R2G	1st Line: V8560 2nd Line: 150	1.5 V		
NCV8560MN180R2G	1st Line: V8560 2nd Line: 180	1.8 V		
NCV8560MN250R2G	1st Line: V8560 2nd Line: 250	2.5 V		
NCV8560MN280R2G	1st Line: V8560 2nd Line: 280	2.8 V	DFN6 (3x3) (Pb-Free)	3000/Tape & Reel
NCV8560MN300R2G	1st Line: V8560 2nd Line: 300	3.0 V		
NCV8560MN330R2G	1st Line: V8560 2nd Line: 330	3.3 V		
NCV8560MN350R2G	1st Line: V8560 2nd Line: 350	3.5 V		
NCV8560MN500R2G	1st Line: V8560 2nd Line: 500	5.0 V		
NCV8560MN130R2G	1st Line: AQ(M)	1.3 V	DFN6 (2x2.2) (Pb-Free)	3000/Tape & Reel
NCV8560SNADJT1G	LJ9	ADJ		
NCV8560SN130T1G	LJ2	1.3 V		
NCV8560SN150T1G	AAJ	1.5 V		
NCV8560SN180T1G	LJ3	1.8 V		
NCV8560SN250T1G	AAQ	2.5 V	TSOP-5	0000/T 0 DI
NCV8560SN280T1G	AAR	2.8 V	(Pb-Free)	3000/Tape & Reel
NCV8560SN300T1G	LJ4	3.0 V		
NCV8560SN330T1G	LJ5	3.3 V		
NCV8560SN350T1G	LJ7	3.5 V		
NCV8560SN500T1G	LJ8	5.0 V		

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.
*NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP Capable

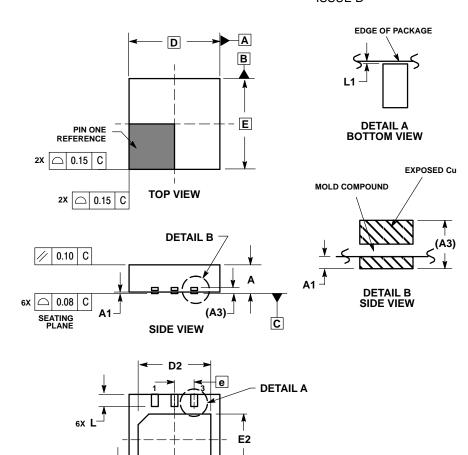
PACKAGE DIMENSIONS



^{*}For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

PACKAGE DIMENSIONS

DFN6 3x3 CASE 488AE **ISSUE B**



6X b NOTE 3

0.05 C

Ф

0.10 C A B

П

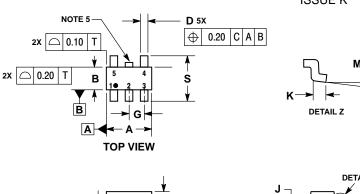
BOTTOM VIEW

- NOTES:
 1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M, 1994.
 2. CONTROLLING DIMENSION: MILLIMETERS.
 3. DIMENSION & APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.25 AND 0.30 MM FROM TERMINAL.
 4. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.
 5. TERMINAL & MAY HAVE MOLD COMPOUND MATERIAL ALONG SIDE EDGE. MOLD FLASHING MAY NOT EXCEED 30 MICRONS ONTO BOTTOM SURFACE OF TERMINAL b.

	MILLIMETERS				
DIM	MIN	MAX			
Α	0.80	1.00			
A1	0.00	0.05			
A3	0.20	0.25			
b	0.18	0.30			
D	3.00 BSC				
D2	2.25	2.55			
Е	3.00	BSC			
E2	1.55	1.85			
е	0.65	BSC			
K	0.20				
L	0.30	0.50			
L1	0.00	0.021			

PACKAGE DIMENSIONS

TSOP-5 CASE 483-02 ISSUE K







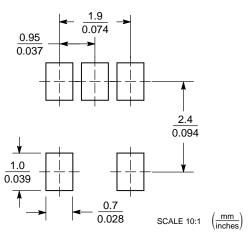
NOTES:

- 1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.

- Y14.5M, 1994.
 CONTROLLING DIMENSION: MILLIMETERS.
 MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH
 THICKNESS. MINIMUM LEAD THICKNESS IS THE
 MINIMUM THICKNESS OF BASE MATERIAL.
 DIMENSIONS A AND B DO NOT INCLUDE MOLD
 FLASH, PROTRUSIONS, OR GATE BURRS. MOLD
 FLASH, PROTRUSIONS, OR GATE BURRS SHALL NOT
 EXCEED 0.15 PER SIDE. DIMENSION A.
 ODITIONAL CONSTRUCTION. AN ARRESTANT
- OPTIONAL CONSTRUCTION: AN ADDITIONAL TRIMMED LEAD IS ALLOWED IN THIS LOCATION.
 TRIMMED LEAD NOT TO EXTEND MORE THAN 0.2 FROM BODY.

	MILLIMETERS					
DIM	MIN	MAX				
Α	3.00	BSC				
В	1.50	BSC				
С	0.90	1.10				
D	0.25	0.50				
G	0.95	BSC				
Н	0.01	0.10				
J	0.10	0.26				
K	0.20 0.60					
M	0° 10°					
S	2.50	3.00				

SOLDERING FOOTPRINT*



*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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