

ZHCS466A - AUGUST 2011 - REVISED OCTOBER 2012

具有

可选增益和自动恢复短路保护的 3.2W 单声道 D 类音频功率放大器

查询样品: TPA2038D1

特性

- 无滤波器单声道 D 类扬声器放大器
- 6dB 和 12dB 间的增益引脚选择
- 5V 电源(10% 总谐波失真 + 噪声(THD + N)) 提供 3.2W 功率进入 4Ω 负载
- 功能强大的单声道 D 类扬声器放大器
 - 在 5V 电源提供 1.4W 功率进入 8Ω 时为 1%
 - 在 5V 电源提供 2.5W 功率进入 4Ω 时为 1%
- 集成图像抑制滤波器用于 DAC 降噪
- 20µV 低输出噪声
- 1.5mA 低静态电流
- 自动恢复短路保护
- 热过载保护
- 9 焊球, 1.21mm x 1.16mm 0.4mm 焊球间距晶圆 级芯片封装 (WCSP)

应用范围

- 无线或者蜂窝手持电话和个人数据助理(PDA)
- 便携式导航设备
- 普通便携式音频器件

说明

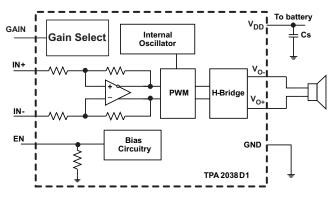
TPA2038D1 是一款 3.2W 功率进入 4 欧姆负载 (10% THD) 高效无滤波器 D 类音频功率放大器。 增益引脚 将增益设定为 6dB 或 12dB。

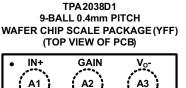
诸如 95% 的效率,1.5mA 静态电流,0.5μA 关断电流,81dB 电源抑制比 (PSRR),20μV 输出噪声和改进的射频 (RF) 抗扰度使得 TPA2038D1 D 类放大器非常适合于蜂窝手机应用。无接通爆音的 4ms 内启动时间。

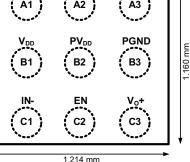
TPA2038D1 采用 1.21mm x 1.16mm, 0.4 焊球间距 晶圆级芯片封装 (WCSP)。

应用电路

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

T _A	PACKAGED DEVICES ⁽¹⁾	PART NUMBER ⁽²⁾	SYMBOL
-40°C to 85°C		TPA2038D1YFFR	QWK
	9-ball WCSP	TPA2038D1YFFT	QWK

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI Web site at www.ti.com

(2) The YFF package is only available taped and reeled. The suffix "R" indicates a reel of 3000, the suffix "T" indicates a reel of 250.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range, $T_A = 25^{\circ}C$ (unless otherwise noted)⁽¹⁾

			VALUE	UNIT
	Cupply voltogo	In active mode	-0.3 to 6.0	V
V _I R _L	Supply voltage	In shutdown mode	-0.3 to 6.0	V
VI	Input voltage	EN, IN+, IN-	–0.3 to V _{DD} + 0.3	V
R _L	Minimum load resistance		3.2	Ω
	Output continuou	is total power dissipation	See Dissipation Rating Table	
T _A	Operating free-a	r temperature range	-40 to 85	°C
TJ	Operating junction	on temperature range	-40 to 150	°C
T _{stg}	Storage tempera	ture range	-65 to 85	°C

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods may affect device reliability.

DISSIPATION RATINGS

PACKAGE	DERATING FACTOR ⁽¹⁾	T _A < 25°C	T _A = 70°C	T _A = 85°C
YFF (WCSP)	4.2 mW/°C	525 mW	336 mW	273 mW

(1) Derating factor measure with high K board.

RECOMMENDED OPERATING CONDITIONS

			MIN	MAX	UNIT
V _{DD} , PV _{DD}	Class-D supply voltage	2.5	5.5	V	
VIH	High-level input voltage	EN, GAIN	1.3		V
VIL	Low-level input voltage	EN, GAIN		0.35	V
VIC	Common mode input voltage range	V _{DD} = 2.5 V, 5.5 V, CMRR ≥ 49 dB	0.75	V _{DD} -1.1	V
T _A	Operating free-air temperature		-40	85	°C

GAIN SETTING

GAIN PIN	GAIN SETTING
GND	12 dB
VDD	6 dB





ELECTRICAL CHARACTERISTICS

 $PVDD = VDD = 3.6 V, T_A = 25^{\circ}C$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{os}	Output offset voltage (measured differentially)	VDD = 2.5 V to 5.5 V, GAIN = VDD		1	5	mV
Іін	High-level EN input current	VDD = 5.5 V, EN = GAIN = 5.5 V			50	μA
I _{IL}	Low-level EN input current	VDD = 5.5 V, EN = GAIN = 0 V			1	μA
		VDD = 5.5 V, no load		1.8	2.5	
I _(Q)	Quiescent current	VDD = 3.6 V, no load		1.5	2.3	mA
		VDD = 2.5 V, no load		1.3	2.1	
I _(SD)	Shutdown current	EN = 0.35 V, VDD = 3.6 V		0.1	2	μA
R _{O, SD}	Output impedance in shutdown mode	EN = 0.35 V		2		kΩ
f _(SW)	Switching frequency	VDD = 2.5 V to 5.5 V	250	300	350	kHz
A _V	Gain	GAIN = 0 V	11.5	12	12.5	
		GAIN = VDD	5.5	6	6.5	dB
R _{EN}	Resistance from EN to GND			300		kΩ
		$A_V = 6 \text{ dB}; \text{EN} = \text{VDD}$		150		
R _{IN}	Single-ended input resistance	$A_V = 12 \text{ dB}; \text{EN} = \text{VDD}$		75		kΩ
		EN = 0.35 V		75		

OPERATING CHARACTERISTICS

 $\mathsf{PVDD}=\mathsf{VDD}=3.6$ V, $\mathsf{A}_\mathsf{V}=6$ dB, $\mathsf{T}_\mathsf{A}=25^\circ\mathsf{C},\,\mathsf{R}_\mathsf{L}=8$ Ω (unless otherwise noted)

	PARAMETER	TEST CONDITIC	NS	MIN TYP	MAX	UNIT
			$V_{DD} = 5 V$	3.24		
		THD + N = 10%, f = 1 kHz, R ₁ = 4 Ω	V _{DD} = 3.6 V	1.62		W
			V _{DD} = 2.5 V	0.70		
			$V_{DD} = 5 V$	2.57		
		THD + N = 1%, f = 1 kHz, R ₁ = 4 Ω	$V_{DD} = 3.6 V$	1.32		W
P _O	Output power		$V_{DD} = 2.5 V$	0.57		
	Output power		$V_{DD} = 5 V$	1.80		
		THD + N = 10%, f = 1 kHz, R _I = 8 Ω	$V_{DD} = 3.6 V$	0.91		w
			$V_{DD} = 2.5 V$	0.42		
		THD + N = 1%, f = 1 kHz, $R_L = 8 \Omega$	$V_{DD} = 5 V$	1.46		
			$V_{DD} = 3.6 V$	0.74		
			$V_{DD} = 2.5 V$	0.33		
	Output voltage noise, $A_V = 6 \text{ dB}$	V_{DD} = 3.6 V, Inputs AC grounded with C _I = 2 µF, f = 20 Hz to 20 kHz	A-weighting	20		μV _{RMS}
E _N			No weighting	26		
∟N			A-weighting	27		
	Output voltage noise, $A_V = 12 \text{ dB}$		No weighting	36		
		$VDD = 5.0 V, P_0 = 1.0 W, f =$	1 kHz, $R_L = 8 \Omega$	0.12%		
		VDD = 3.6 V, P_0 = 0.5 W, f =	1 kHz, $R_L = 8 \Omega$	0.05%		
THD+N	Total harmonic distortion plus noise	VDD = 2.5 V, P_0 = 0.2 W, f = 1 kHz, R_L = 8 Ω		0.05%		
	Total narmonic distortion plus noise	$VDD = 5.0 V, P_0 = 2.0 W, f =$	1 kHz, $R_L = 4 \Omega$	0.32%		
		$VDD = 3.6 V, P_0 = 1.0 W, f =$	1 kHz, $R_L = 4 \Omega$	0.11%		
		VDD = 2.5 V, P_O = 0.4 W, f = 1 kHz, R_L = 4 Ω		0.12%		
DEDD		$A_V = 6 \text{ dB}$, Inputs AC grounde 200 mV _{pp} ripple, f = 217 Hz	d with $C_1 = 2 \mu F$,	81		dP
PSRR	AC power supply rejection ratio	A_V = 12 dB, Inputs AC ground C _I = 2 µF, 200 mV _{pp} ripple, f =	ed with 217 Hz	82		dB

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OPERATING CHARACTERISTICS (continued)

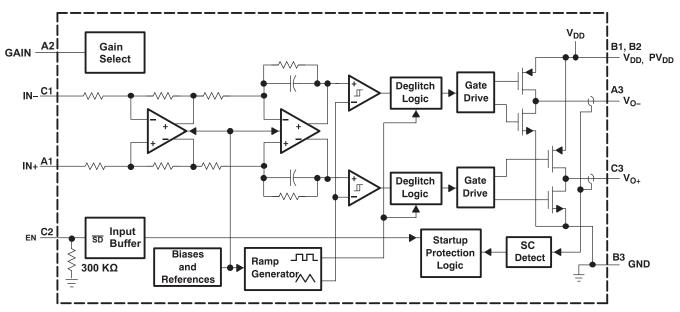
PVDD = VDD = 3.6 V, $A_V = 6 \text{ dB}$, $T_A = 25^{\circ}\text{C}$, $R_L = 8 \Omega$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
CMRR	Common mode rejection ratio	$A_V = 6 \text{ dB}, V_{IC} = 200 \text{ mV}_{PP}, \text{ f} = 217 \text{ Hz}$		79		dB
		$A_V = 12 \text{ dB}, V_{IC} = 200 \text{ mV}_{PP}, \text{ f} = 217 \text{ Hz}$		77		uБ
T _{SU}	Startup time from shutdown			4		ms
		V _{O+} shorted to VDD				
		V _{O-} shorted to VDD				
I _{SC}	Short circuit protection threshold	V _{O+} shorted to GND		2		А
		V _{O-} shorted to GND				
		V_{O+} shorted to V_{O-}				
T _{AR}	Overcurrent recovery time	VDD = 2.5 V to 5.5 V		100		ms

Terminal Functions

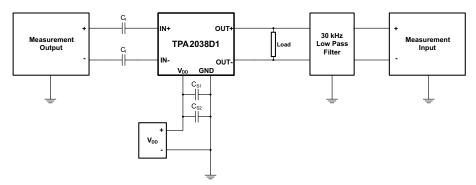
TEI	TERMINAL		DESCRIPTION	
NAME	WCSP BALL	I/O	DESCRIPTION	
IN+	A1	Ι	ositive audio input.	
GAIN	A2	Ι	Gain select. Set to GND for 12 dB; set to VDD for 6 dB.	
V _{O-}	A3	0	Negative audio output.	
VDD	B1	Ι	Power supply terminal. Connect to PVDD using a direct connection.	
PVDD	B2	Ι	Class-D output power supply. Connect to VDD using a direct connection.	
GND	B3	Ι	Ground.	
IN-	C1	Ι	Negative audio input.	
EN	C2	Ι	Enable. Set to logic high to enable device.	
V _{O+}	C3	0	Positive audio output.	

FUNCTIONAL BLOCK DIAGRAM





TEST SETUP FOR GRAPHS



- 1. C_1 was shorted for any common-mode input voltage measurement. All other measurements were taken with $C_1 = 0.1 \ \mu F$ (unless otherwise noted).
- 2. $C_{S1} = 0.1 \ \mu\text{F}$ is placed very close to the device. The optional $C_{S2} = 10 \ \mu\text{F}$ is used for datasheet graphs.
- 3. The 30 kHz low-pass filter is required even if the analyzer has an internal low-pass filter. An RC low-pass filter (1 k Ω , 4700 pF) is used on each output for the data sheet graphs.

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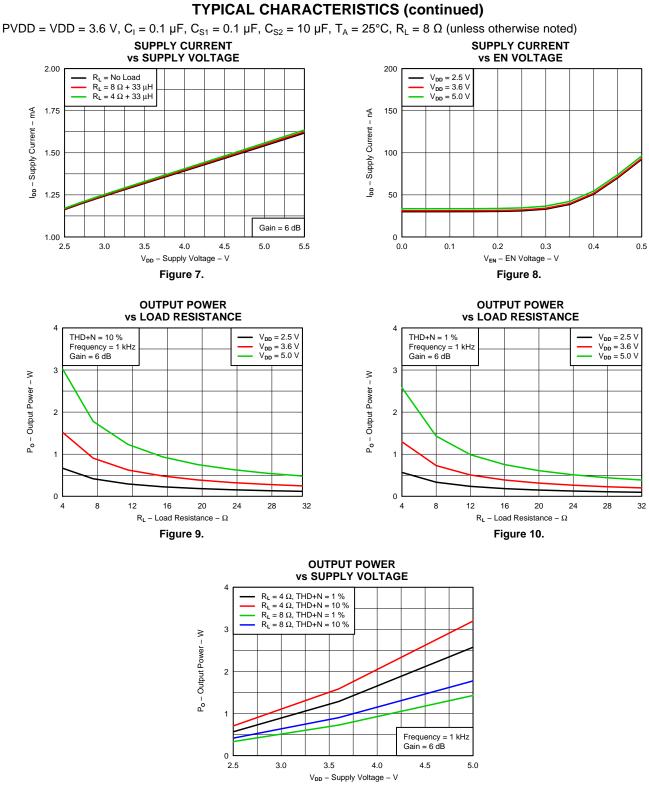
TYPICAL CHARACTERISTICS $PVDD = VDD = 3.6 \text{ V}, \text{ } \text{C}_{\text{I}} = 0.1 \text{ } \mu\text{F}, \text{ } \text{C}_{\text{S1}} = 0.1 \text{ } \mu\text{F}, \text{ } \text{C}_{\text{S2}} = 10 \text{ } \mu\text{F}, \text{ } \text{T}_{\text{A}} = 25^{\circ}\text{C}, \text{ } \text{R}_{\text{L}} = 8 \text{ } \Omega \text{ (unless otherwise noted)}$ EFFICIENCY EFFICIENCY vs OUTPUT POWER vs OUTPUT POWER 100 100 90 90 80 80 70 70 % % 60 60 η – Efficiency η – Efficiency 50 50 40 40 $\mathsf{R}_\mathsf{L}=8~\Omega+33~\mu\mathsf{H}$ $R_L = 4 \Omega + 33 \mu H$ 30 30 Gain = 6 dB Gain = 6 dB 20 20 $V_{DD} = 2.5 V$ $V_{DD} = 3.3 V$ $V_{DD} = 2.5 V$ $V_{DD} = 3.3 V$ 10 10 V_{DD} = 5.0 V V_{DD} = 5.0 V 0 0 0.0 0.2 0.4 0.6 0.8 1.0 1.2 1.4 1.6 1.8 2.0 0.0 0.2 0.4 0.6 0.8 1.0 1.2 1.4 1.6 1.8 2.0 Po - Output Power - W Po - Output Power - W Figure 1. Figure 2. POWER DISSIPATION POWER DISSIPATION vs OUTPUT POWER vs OUTPUT POWER 0.5 0.6 V_{DD} = 3.6 V V_{DD} = 5.0 V $R_L = 8 \Omega + 33 \mu H$ $R_L = 8 \Omega + 33 \mu H$ $R_{L} = 4 \Omega + 33 \mu H$ $R_1 = 4 \Omega + 33 \mu H$ Gain = 6 dBGain = 6 dB0.5 0.4 \ge ≥__ Power Dissipation 0.4 - Power Dissipation 0.3 0.3 0.2 0.2 Å å 0.1 0.1 0.0 0.0 0.0 0.4 0.8 1.2 1.6 2.0 0.0 0.4 0.8 1.2 1.6 2.0 2.4 2.8 3.2 3.6 4.0 Po - Output Power - W Po - Output Power - W Figure 3. Figure 4. SUPPLY CURRENT SUPPLY CURRENT vs OUTPUT POWER vs OUTPUT POWER 1 600m V_{DD} = 2.5 V V_{DD} = 2.5 V 900m V_{DD} = 3.6 V V_{DD} = 3.6 V V_{DD} = 5.0 V 500m V_{DD} = 5.0 V 800m ۹ -Supply Current – A 700m 400m Supply Current 600m 500m 300m 400m 200m 300m B 200m 100m $R_{L} = 4 \Omega + 33 \mu H$ $R_{L} = 8 \Omega + 33 \mu H$ 100m Gain = 6 dB Gain = 6 dB 0 0 0.0 0.2 0.4 0.6 0.8 1.0 1.2 1.4 1.6 1.8 2.0 2.2 2.4 2.6 2.8 3.0 3.2 3.4 3.6 0.4 0.6 0.8 1.0 1.2 1.6 1.8 2.0 0.0 0.2 1.4 Po - Output Power - W Po - Output Power - W Figure 5. Figure 6.

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TPA2038D1

100

10

1

0.1

0.01

10 % -

1

0.1

0.01

0.001

10

1

0.1

0.01

0.001

20

20

Total Harmonic Distortion + Noise

THD+N -

%

THD+N – Total Harmonic Distortion + Noise –

8

10m

%

Harmonic Distortion + Noise

THD+N - Total

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THD + NOISE THD + NOISE vs OUTPUT POWER (6 dB GAIN) vs OUTPUT POWER (6 dB GAIN) 100 % $R_{L} = 4 \Omega + 33 \mu H$ R_L = 8 Ω + 33 μH V_{DD} = 2.5 V V_{DD} = 2.5 V V_{DD} = 3.6 V Gain = 6 dB V_{DD} = 3.6 V Gain = 6 dB THD+N – Total Harmonic Distortion + Noise V_{DD} = 5.0 V V_{DD} = 5.0 V 10 1 0.1 0.01 100m 10m 100m 1 5 1 Po - Output Power - W Po - Output Power - W Figure 12. Figure 13. THD + NOISE THD + NOISE vs FREQUENCY (6 dB GAIN) vs FREQUENCY (6 dB GAIN) 10 THD+N – Total Harmonic Distortion + Noise – % V_{DD} = 5.0 V $P_0 = 50 \text{ mW}$ V_{DD} = 3.6 V Po = 25 mW Po = 125 mW $R_1 = 8 \Omega + 33 \mu H$ $P_0 = 250 \text{ mW}$ $R_1 = 8 \Omega + 33 \mu H$ $P_0 = 1 W$ $P_0 = 500 \text{ mW}$ Gain = 6 dBGain = 6 dB 1 0.1 0.01 0.001 100 10k 20k 20 100 20k 1k 1k 10k f - Frequency - Hz f - Frequency - Hz Figure 14. Figure 15. THD + NOISE THD + NOISE vs FREQUENCY (6 dB GAIN) vs FREQUENCY (6 dB GAIN) 10 % V_{DD} = 2.5 V R_L = 8 Ω + 33 μH $P_0 = 15 \text{ mW}$ $P_0 = 75 \text{ mW}$ $V_{DD} = 5.0 V$ $R_L = 4 \Omega + 33 \mu H$ $P_o = 100 \text{ mW}$ $P_o = 500 \text{ mW}$ THD+N – Total Harmonic Distortion + Noise – $P_0 = 200 \text{ mW}$ $P_0 = 2 W$ Gain = 6 dB Gain = 6 dB 1 0.1 # 0.01 +++ +0.001 100 1k 10k 20k 20 100 1k 10k 20k f - Frequency - Hz f - Frequency - Hz Figure 16. Figure 17. Copyright © 2011–2012, Texas Instruments Incorporated

TYPICAL CHARACTERISTICS (continued)

 $PVDD = VDD = 3.6 \text{ V}, \text{ } \text{C}_{\text{I}} = 0.1 \text{ } \mu\text{F}, \text{ } \text{C}_{\text{S1}} = 0.1 \text{ } \mu\text{F}, \text{ } \text{C}_{\text{S2}} = 10 \text{ } \mu\text{F}, \text{ } \text{T}_{\text{A}} = 25^{\circ}\text{C}, \text{ } \text{R}_{\text{L}} = 8 \text{ } \Omega \text{ (unless otherwise noted)}$

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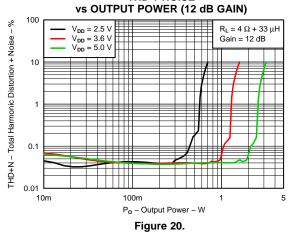
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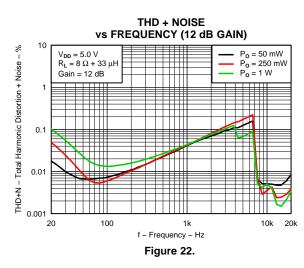
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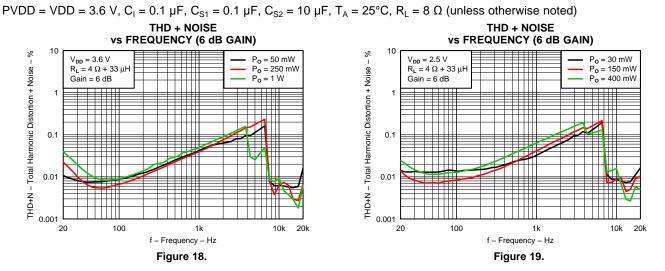
THD + NOISE vs FREQUENCY (6 dB GAIN) 10 % $V_{DD} = 3.6 V$ $R_L = 4 \Omega + 33 \mu h$ $P_0 = 50 \text{ mW}$ $P_0 = 250 \text{ mW}$ THD+N – Total Harmonic Distortion + Noise Gain = 6 dB $P_0 = 1 W$ 1 0.1 0.01 0.001 20 100 1k 10k 20k f - Frequency - Hz Figure 18.

TYPICAL CHARACTERISTICS (continued)

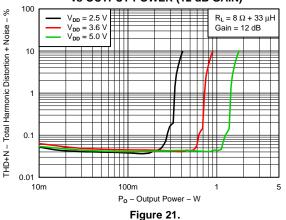


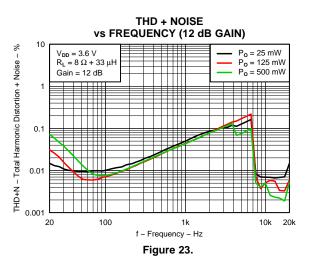






THD + NOISE vs OUTPUT POWER (12 dB GAIN)

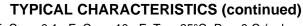


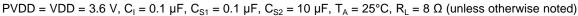


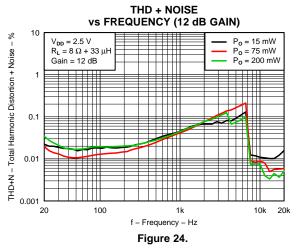
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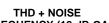
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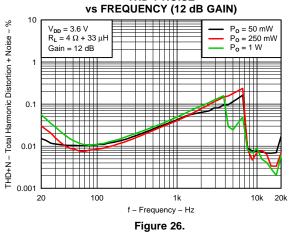
THD + NOISE vs FREQUENCY (12 dB GAIN) 10 V_{DD} = 5.0 V R_L = 4 Ω + 33 μł $P_0 = 100 \text{ mW}$ $P_0 = 500 \text{ mW}$ Gain = 12 dB $P_0 = 2 W$ 1

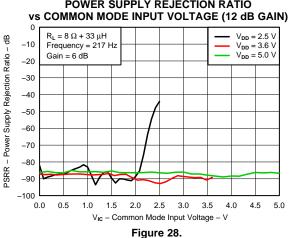




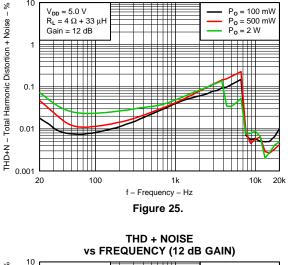


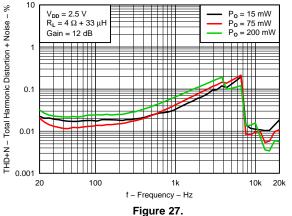




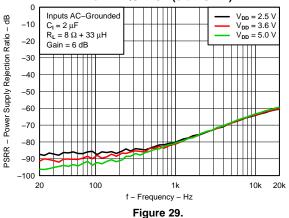












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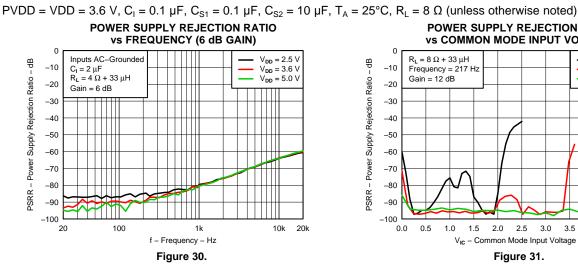
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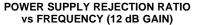


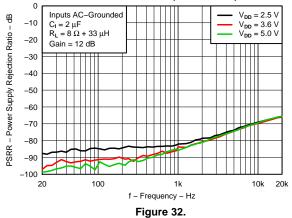
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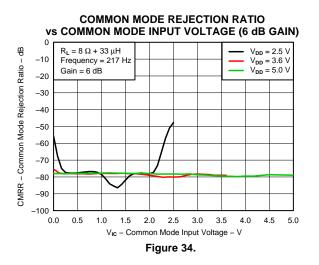
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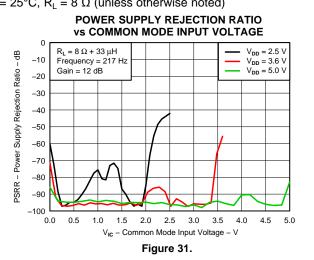
TYPICAL CHARACTERISTICS (continued)



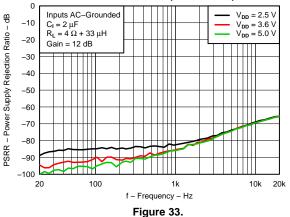




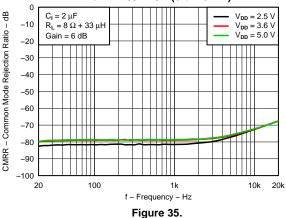




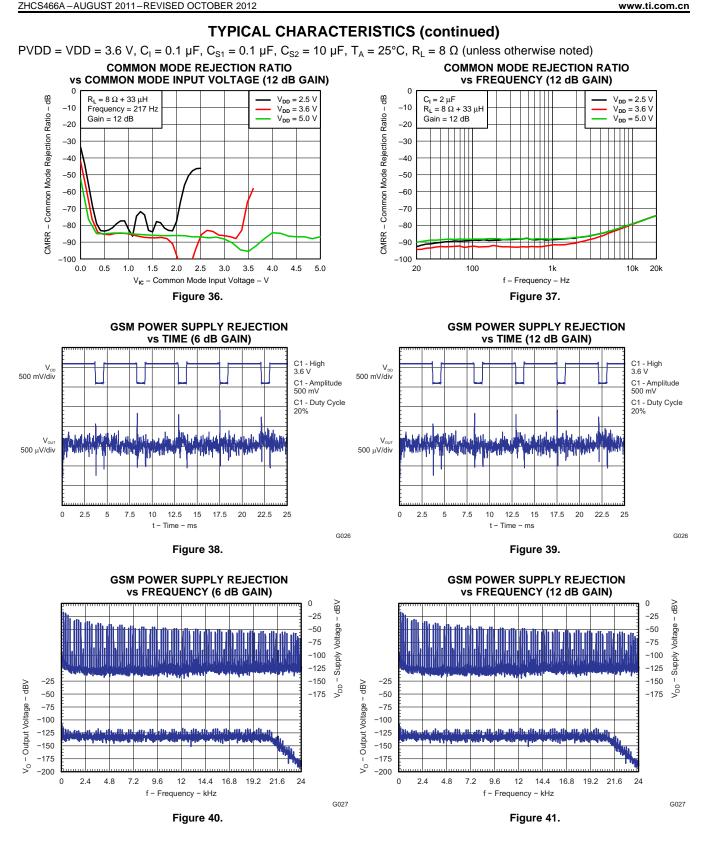
POWER SUPPLY REJECTION RATIO vs FREQUENCY (12 dB GAIN)







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INSTRUMENTS



APPLICATION INFORMATION

SHORT CIRCUIT AUTO-RECOVERY

When a short-circuit event occurs, the TPA2038D1 goes to shutdown mode and activates the integrated autorecovery process whose aim is to return the device to normal operation once the short-circuit is removed. This process repeatedly examines (once every 100 ms) whether the short-circuit condition persists, and returns the device to normal operation immediately after the short-circuit condition is removed. This feature helps protect the device from large currents and maintain a good long-term reliability.

INTEGRATED IMAGE REJECT FILTER FOR DAC NOISE REJECTION

In applications which use a DAC to drive Class-D amplifiers, out-of-band noise energy present at the DAC's image frequencies fold back into the audio-band at the output of the Class-D amplifier. An external low-pass filter is often placed between the DAC and the Class-D amplifier in order to attenuate this noise.

The TPA2038D1 has an integrated Image Reject Filter with a low-pass cutoff frequency of 130 kHz, which significantly attenuates this noise. Depending on the system noise specification, the integrated Image Reject Filter may help eliminate external filtering, thereby saving board space and component cost.

COMPONENT SELECTION

Figure 42 shows the TPA2038D1 typical schematic with differential inputs, while Figure 43 shows the TPA2038D1 with differential inputs and input capacitors. Figure 44 shows the TPA2038D1 with a single-ended input.

Decoupling Capacitors (C_{S1}, C_{S2})

The TPA2038D1 is a high-performance class-D audio amplifier that requires adequate power supply decoupling to ensure the efficiency is high and total harmonic distortion (THD) is low. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor $C_{S1} = 0.1 \,\mu\text{F}$, placed as close as possible to the device V_{DD} lead works best. Placing C_{S1} close to the TPA2038D1 is important for the efficiency of the class-D amplifier, because any resistance or inductance in the trace between the device and the capacitor can cause a loss in efficiency. For filtering lower-frequency noise signals, a 10 μ F or greater capacitor (C_{S2}) placed near the audio power amplifier would also help, but it is not required in most applications because of the high PSRR of this device. Typically, the smaller the capacitor's case size, the lower the inductance and the closer it can be placed to the TPA2038D1. X5R and X7R dielectric capacitors are recommended for both C_{S1} and C_{S2} .

Input Capacitors (C_I)

The TPA2038D1 does not require input coupling capacitors if the design uses a differential source that is biased within the common-mode input voltage range. That voltage range is listed in the Recommended Operating Conditions table. If the input signal is not biased within the recommended common-mode input range, such as in needing to use the input as a high pass filter, shown in Figure 43, or if using a single-ended source, shown in Figure 44, input coupling capacitors are required. The same value capacitors should be used on both IN+ and IN- for best pop performance. The 3 dB high-pass cutoff frequency f_c of the filter formed by the input coupling capacitor C_1 and the input resistance R_1 (typically 150 k Ω) of the TPA2038D1 is given by Equation 1:

$$f_{\rm C} = \frac{1}{\left(2\pi R_{\rm I} C_{\rm I}\right)} \tag{1}$$

The value of the input capacitor is important to consider as it directly affects the bass (low frequency) performance of the circuit. Speaker response may also be taken into consideration when setting the corner frequency using input capacitors. Solving for the input coupling capacitance, we get:

$$C_{I} = \frac{1}{\left(2\pi R_{I} f_{C}\right)} \tag{2}$$

If the corner frequency is within the audio band, the capacitors should have a tolerance of $\pm 10\%$ or better, because any mismatch in capacitance causes an impedance mismatch at the corner frequency and below.

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For a flat low-frequency response, use large input coupling capacitors (0.1 µF or larger). X5R and X7R dielectric capacitors are recommended.

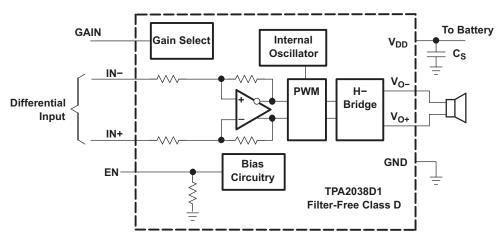


Figure 42. Typical TPA2038D1 Application Schematic With DC-coupled Differential Input

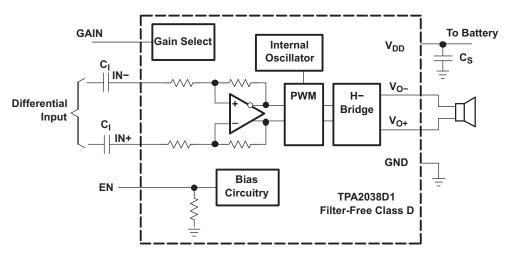


Figure 43. TPA2038D1 Application Schematic With Differential Input and Input Capacitors

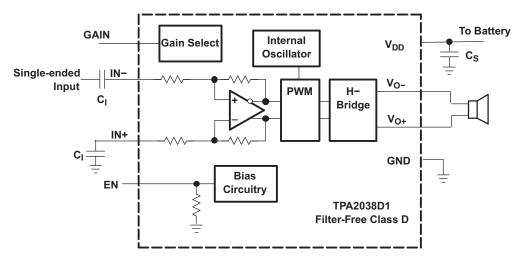


Figure 44. TPA2038D1 Application Schematic With Single-Ended Input



EFFICIENCY AND THERMAL INFORMATION

The maximum ambient operating temperature of the TPA2038D1 depends on the load resistance, power supply voltage and heat-sinking ability of the PCB system. The derating factor for the YFF package is shown in the dissipation rating table. Converting this to θ_{JA} :

$$\theta_{\rm JA} = \frac{1}{\rm Derating \ Factor}$$
(3)

Given θ_{JA} (from the Package Dissipation ratings table), the maximum allowable junction temperature (from the Absolute Maximum ratings table), and the maximum internal dissipation (from Power Dissipation vs Output Power figures) the maximum ambient temperature can be calculated with the following equation. Note that the units on these figures are Watts RMS. Because audio contains crest factors (ratio of peak power to RMS power) from 9–15 dB, thermal limitations are not usually encountered.

$$T_A^{Max} = T_J^{Max} - \theta_{JA}^P D_{max}$$
(4)

The TPA2038D1 is designed with thermal protection that turns the device off when the junction temperature surpasses 150°C to prevent damage to the IC. Note that the use of speakers less resistive than 4 Ω (typ) is not advisable. Below 4 Ω (typ) the thermal performance of the device dramatically reduces because of increased output current and reduced amplifier efficiency. The Absolute Maximum rating of 3.2 Ω covers the manufacturing tolerance of a 4 Ω speaker and speaker impedance decrease due to frequency. θ_{JA} is a gross approximation of the complex thermal transfer mechanisms between the device and its ambient environment. If the θ_{JA} calculation reveals a potential problem, a more accurate estimate should be made.

WHEN TO USE AN OUTPUT FILTER

Design the TPA2038D1 without an Inductor / Capacitor (LC) output filter if the traces from the amplifier to the speaker are short. Wireless handsets and PDAs are great applications for this class-D amplifier to be used without an output filter.

The TPA2038D1 does not require an LC output filter for short speaker connections (approximately 100 mm long or less). A ferrite bead can often be used in the design if failing radiated emissions testing without an LC filter; and, the frequency-sensitive circuit is greater than 1 MHz. If choosing a ferrite bead, choose one with high impedance at high frequencies, but very low impedance at low frequencies. The selection must also take into account the currents flowing through the ferrite bead. Ferrites can begin to loose effectiveness at much lower than rated current values. See the EVM User's Guide (SLOU298) for components used successfully by TI.

Figure 45 shows a typical ferrite-bead output filter.

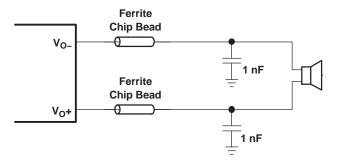


Figure 45. Typical Ferrite Chip Bead Filter

PRINTED CIRCUIT BOARD LAYOUT

In making the pad size for the WCSP balls, it is recommended that the layout use non-solder-mask-defined (NSMD) land. With this method, the solder mask opening is made larger than the desired land area, and the opening size is defined by the copper pad width. Figure 46 shows the appropriate diameters for a WCSP layout.

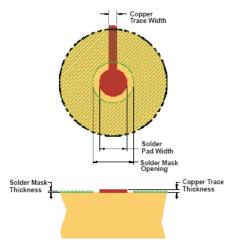


Figure 46. Land Pattern Image and Dimensions

SOLDER PAD DEFINITIONS	COPPER PAD	SOLDER MASK OPENING ⁽⁵⁾	COPPER THICKNESS	STENCIL OPENING ⁽⁶⁾⁽⁷⁾	STENCIL THICKNESS
Non-solder-mask- defined (NSMD)	0.23 mm	0.310 mm	1 oz max (0.032 mm)	0.275 mm x 0.275 mm Sq. (rounded corners)	0.1 mm thick

- 1. Circuit traces from NSMD defined PWB lands should be 75 μm to 100 μm wide in the exposed area inside the solder mask opening. Wider trace widths reduce device stand off and impact reliability.
- 2. Best reliability results are achieved when the PWB laminate glass transition temperature is above the operating the range of the intended application.
- 3. Recommend solder paste is Type 3 or Type 4.
- 4. For a PWB using a Ni/Au surface finish, the gold thickness should be less 0.5 mm to avoid a reduction in thermal fatigue performance.
- 5. Solder mask thickness should be less than 20 µm on top of the copper circuit pattern
- 6. Best solder stencil performance is achieved using laser cut stencils with electro polishing. Use of chemically etched stencils give inferior solder paste volume control.
- 7. Trace routing away from WCSP device should be balanced in X and Y directions to avoid unintentional component movement due to solder wetting forces.

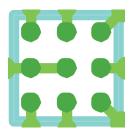


Figure 47. Layout Snapshot

An on-pad via is not required to route the middle ball B2 (PV_{DD}) of the TPA2038D1. Short ball B2 (PV_{DD}) to ball B1 (V_{DD}) and connect both to the supply trace as shown in Figure 47. This simplifies board routing and saves manufacturing cost.



PACKAGE DIMENSIONS

D	E
Max = 1190µm	Max = 1244µm
Min = 1130µm	Min = 1184µm



REVISION HISTORY

Cł	Changes from Original (August 2011) to Revision A Page						
•	Changed 说明的第一句,8 欧姆至 4 欧姆	1					
•	Changed D and E dimensions in the Package Dimensions table.	17					



10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPA2038D1YFFR	ACTIVE	DSBGA	YFF	9	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	QWK	Samples
TPA2038D1YFFT	ACTIVE	DSBGA	YFF	9	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	QWK	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <= 1000ppm threshold. Antimony trioxide based flame retardants must also meet the <= 1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

10-Dec-2020

PACKAGE MATERIALS INFORMATION

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Texas Instruments

TAPE AND REEL INFORMATION



*All dimensions are nominal



QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPA2038D1YFFR	DSBGA	YFF	9	3000	180.0	8.4	1.34	1.34	0.81	4.0	8.0	Q1
TPA2038D1YFFT	DSBGA	YFF	9	250	180.0	8.4	1.34	1.34	0.81	4.0	8.0	Q1

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PACKAGE MATERIALS INFORMATION

18-Jun-2020



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPA2038D1YFFR	DSBGA	YFF	9	3000	182.0	182.0	20.0
TPA2038D1YFFT	DSBGA	YFF	9	250	182.0	182.0	20.0

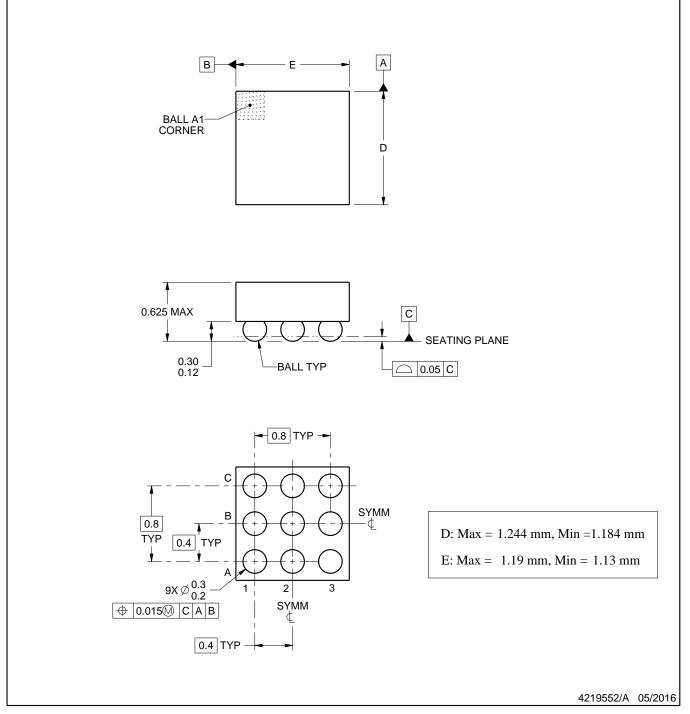
YFF0009



PACKAGE OUTLINE

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.2. This drawing is subject to change without notice.

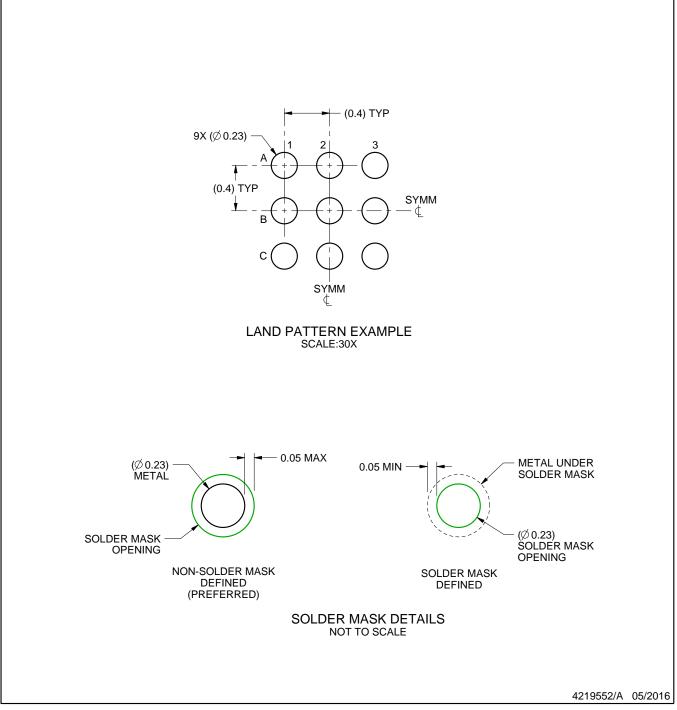


YFF0009

EXAMPLE BOARD LAYOUT

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. For more information, see Texas Instruments literature number SNVA009 (www.ti.com/lit/snva009).

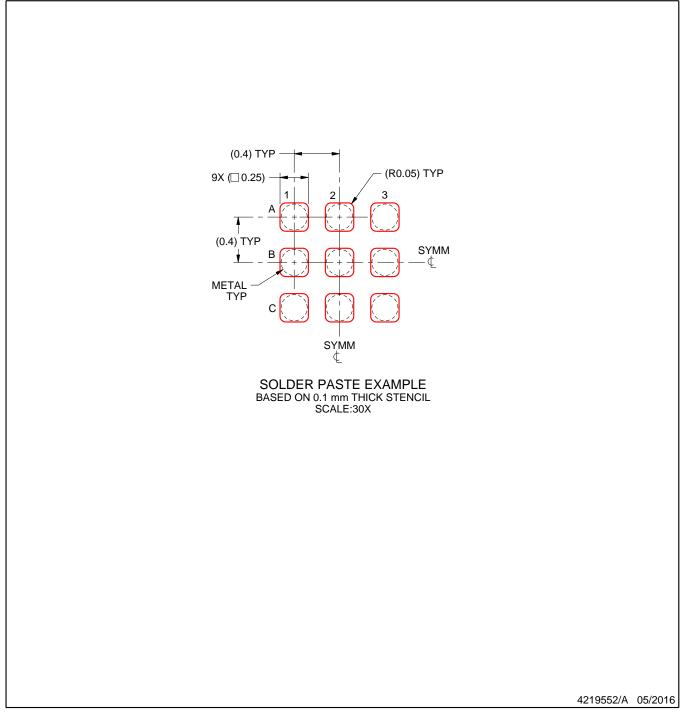


YFF0009

EXAMPLE STENCIL DESIGN

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.



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