# 双通道，高压－多级输出完全集成超声波发射机 <br> 查询样品：TX517 

## 特性

- 输出电压：
- 差分模式下，高达 200Vpp
- 峰值输出电流：$\pm 2.5 \mathrm{~A}$
- 多级输出
- 差分：17级
- 单端：5级
- 集成：
- 电平转换器
- 驱动器
- 高压输出级
- CW 输出
- TX 输出更新率
- 最高 100MSPS
- 最小外部组件
- 小型封装： $13 \times 13 \mathrm{~mm}$ 球状引脚栅格阵列（BGA）封装


## 应用范围

说明
TX517 是一款完全集成，双通道，高压发射器。它特别设计用于高要求的医疗超声波应用，此应用要求一个多级高压脉冲波形。此输出级设计传送典型值为 $\pm 2.5 \mathrm{~A}$的峰值输出电流，摆幅200Vpp。

TX517 是一个完整发射器解决方案，此方案每通道含低压输入逻辑，电平转换器，栅极驱动器和 P 通道， N通道金属氧化物半导体场效应晶体管（MOSFET）。

TX517 还包含一个 CW 输出级。
TX517 采用球状引脚栅格阵列（BGA）封装，此封装是无铅（RoHS兼容）并且绿色环保。它的额定运行温度为 $0^{\circ} \mathrm{C}$ 至 $85^{\circ} \mathrm{C}$ 。

## 17个电平脉冲发生器芯片：

当与变压器一同使用时，此芯片包含 2 个 5 级通道来形成一个单一17级发射器单元。 此器件设计用于驱动处于不同输出电平上的变送器，而且能调制输出脉冲宽度以获得脉宽调制频谱定型的额外的灵活性。

- 医疗超声波
- 高压信号生成器


Please be aware that an important notice concerning availability，standard warranty，and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet．

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## PACKAGING/ORDERING INFORMATION ${ }^{(1)}$

| PACKAGED DEVICES | PACKAGE TYPE | PACKAGE MARKING | TRANSPORT MEDIA, QUANTITY | ECO STATUS ${ }^{(2)}$ |
| :---: | :---: | :---: | :---: | :---: |
| TX517IZCQ | BGA-144 | TX517 | Tray | Pb-Free, Green |

(1) NOTE: These Packages conform to Lead-Free and Green Manufacturing Specifications
(2) Eco-Status information: Additional details including specific material content can be accessed at www.ti.com/leadfree

GREEN: Ti defines Green to mean Lead (Pb)-Free and in addition, uses less package materials that do not contain halogens, including bromine ( Br ), or antimony ( Sb ) above $0.1 \%$ of total product weight.
N/A: Not yet available Lead (Pb)-Free; for estimated conversion dates, go to www.ti.com/leadfree.
Pb-FREE: Ti defines Lead (Pb)-Free to mean RoHS compatible, including a lead concentration that does not exceed $0.1 \%$ of total product weight, and, if designed to be soldered, suitable for use in specified lead-free soldering processes.

DEVICE INFORMATION


## PIN FUNCTIONS

| PIN NAME | DESCRIPTION |
| :---: | :---: |
| SUPPLIES |  |
| VAAx | Input Logic Supply (+2.5V) |
| VDD | +5V Driver Supply |
| VEE | -5V Driver Supply |
| HVOA, HVOB | Positive Supply of Low-voltage FET Output stage; Channel A and B |
| LVOA, LVOB | Negative Supply of Low-voltage FET Output stage; Channel A and B |
| HV2A, HV2B | Positive Supply of Intermediate voltage FET Output stage; this stage includes an internal de-glitcher circuit.Channel A and B |
| LV2A, LV2B | Negative Supply of Intermediate voltage FET Output stage; this stage includes an internal de-glitcher circuit.Channel A and B |
| HV1A, HV1B | Positive Supply of High-voltage FET Output stage; Channel A and B |
| LV1A, LV1B | Negative Supply of High-voltage FET Output stage; Channel A and B |
| VCWA, VCWB | Supply connections for CW FET output stage; Channel A and B |
| GND | Ground connection; Driver |
| GNDCWA, GNDCWB | Ground connection for CW FET output stage of Channel A and B |
| BIAS | Connect to VAA (+2.5V); used for internal biasing; high-impedance input |
| INPUTS |  |
| INPOA, INPOB | Logic input signal for the Low-voltage P-FET stage of channel A and B; Low = ON, High = OFF. Controls HVOA, HVOB. High impedance input. |
| INNOA, INNOB | Logic input signal for the Low-voltage N-FET stage of channel A and B; Low = OFF, High = ON. Controls LVOA, LVOB. High impedance input. |
| INP2A, INP2B | Logic input signal for the Intermediate voltage P-FET stage of channel A and B; Low = ON, High = OFF. Controls HV2A, HV2B. High impedance input. |
| INN2A, INN2B | Logic input signal for the Intermediate Voltage N-FET stage of channel A and B; Low = OFF, High = ON. Controls LV2A, LV2B. High impedance input. |
| INP1A, INP1B | Logic input signal for the High-voltage P-FET stage of channel A and B; Low = ON, High = OFF. Controls HV1A, HV1B. High impedance input. |
| INN1A, INN1B | Logic input signal for the High-voltage N-FET stage of channel A and B; Low = OFF, High = ON. Controls LV1A, LV1B. High impedance input. |
| CWINA | CW gate input signal for A output. An input ' 1 ' means that current sinks from OUTA. An input ' 0 ' means that current sources from OUTA. This pin directly accesses the output A CW FET gates. |
| CWINB | CW gate input signal for B output. An input ' 1 ' means that current sinks from OUTB. An input ' 0 ' means that current sources from OUTB. This pin directly accesses the output B CW FET gates. |
| $\overline{\mathrm{EN}}$ | Logic Input for non-CW path; use the Enable-pin to select between input data being latched or transparent operation. Low = input data will be retimed by the internal $(\mathrm{T} \mathrm{\& H})$ at the rate of the applied clock at PCLKIN. High = use this mode when operating the TX517 without a clock. When High (1) the input data will bypass the $(T \& H)$. This pin is a common control for Channel A and B. High impedance input. |
| $\overline{P D M}$ | Power-down control input non-CW path; Low = power-down, High = normal operation. The $\overline{\text { PDM }}$-pin controls the voltage translation circuits which draw some quiescent power. This pin is a common control for Channel A and B. High impedance input. |
| PCLKIN | Clock input for usage in latch (T\&H) mode. When clock signal is high, the (T\&H) circuit is in track mode. When clock signal is low, the (T\&H) is in hold mode. This pin is a common clock input for both Channel A and B. High impedance input. |
| OUTPUTS |  |
| OUTA | Output Channel A |
| OUTB | Output Channel B |

## ABSOLUTE MAXIMUM RATINGS

Voltages referenced to Ground potential（GND $=0 \mathrm{~V}$ ）；over operating free－air temperature（unless otherwise noted）${ }^{(1)}$

|  |  | VALUE | UNIT |
| :---: | :---: | :---: | :---: |
|  | High－Voltage，Positive Supply HV1，2 referred to OUTA／B，see also Max．delta voltage | -0.3 to +80 | V |
|  | High－Voltage，Positive Supply HVO referred to OUTA／B，see also Max．delta voltage | -0.3 to＋6 | V |
|  | High－Voltage VCWA／B supply referred to GNDCWA／B | －0．3 to＋16 | V |
|  | High－Voltage，Negative Supply LV1，2 referred to OUTA／B，see also Max．delta voltage | -40 to +0.3 | V |
| $V_{\text {DS }}$ | High－Voltage，Negative Supply LV0 referred to OUTA／B，see also Max．delta voltage | －6 to +0.3 | V |
|  | Max．delta voltage：HV1－LV1 and HV2－LV2 | 110 | V |
|  | Max．delta voltage：HV0－LV0 | 12 | V |
| VDD | Driver Supply，positive | -0.3 to +6 | V |
| VEE | Driver Supply，negative | －6 to＋0．3 | V |
| VAA | Logic Supply Voltage | -0.3 to +6 | V |
|  | Logic Inputs（INPx，INNx，EN，$\overline{\text { PDM，PCLKIN，U }}$ ） | -0.3 to +6 | V |
|  | CW inputs（CWINA，CWINB） | －0．3 to＋11 | V |
|  | Peak Solder Temperature ${ }^{(2)}$ | 260 | ${ }^{\circ} \mathrm{C}$ |
| TJ | Maximum junction temperature，any condition ${ }^{(3)}$ | 150 | ${ }^{\circ} \mathrm{C}$ |
| TJ | Maximum junction temperature，continuous operation，long term reliability ${ }^{(4)}$ | 125 | ${ }^{\circ} \mathrm{C}$ |
| Tstg | Storage temperature range | －65 to 150 | ${ }^{\circ} \mathrm{C}$ |
|  | HBM | 500 | V |
| ESD ratings | CDM | 750 | V |
|  | MM | 200 | V |

（1）Stresses above those listed under absolute maximum ratings may cause permanent damage to the device．These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under＂recommended operating conditions＂is not implied．Exposure to absolute maximum rated conditions for extended periods may degrade device reliability．
（2）Device complies with JSTD－020D．
（3）The absolute maximum junction temperature under any condition is limited by the constraints of the silicon process．
（4）The absolute maximum junction temperature for continuous operation is limited by the package constraints．Operation above this temperature may result in reduced reliability and／or lifetime of the device．

## THERMAL INFORMATION

| THERMAL METRIC ${ }^{(1)}$ |  | TX517 | UNITS |
| :---: | :---: | :---: | :---: |
|  |  | BGA（144）（ZCQ）PINS |  |
| $\theta_{\mathrm{JA}}$ | Junction－to－ambient thermal resistance | 28 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\text {JCtop }}$ | Junction－to－case（top）thermal resistance | 3.8 |  |
| $\theta_{J B}$ | Junction－to－board thermal resistance | 11.3 |  |
| $\Psi_{\text {JT }}$ | Junction－to－top characterization parameter | 0.2 |  |
| Power <br> Rating ${ }^{(2)(3)}$ <br> （ $\mathrm{TJ}=125^{\circ} \mathrm{C}$ ） | $\mathrm{TA}=25^{\circ} \mathrm{C}$ | 3.57 | W |
|  | $\mathrm{TA}=85^{\circ} \mathrm{C}$ | 1.47 |  |

（1）有关传统和新的热度量的更多信息，请参阅 IC 封装热度量 应用报告 SPRA953。
（2）This data was taken with the JEDEC High－K test PCB．
（3）Power rating is determined with a junction temperature of $125^{\circ} \mathrm{C}$ ．This is the point where distortion starts to substantially increase and long－term reliability starts to be reduced．Thermal management of the final PCB should strive to keep the junction temperature at or below $125^{\circ} \mathrm{C}$ for best performance and reliability．

## RECOMMENDED OPERATING CONDITIONS

|  | MIN | TYP | MAX |
| :--- | ---: | ---: | ---: |
| UNIT |  |  |  |
| VDD | 2.38 | 2.5 | 3.3 |
| VEE | 4.75 | 5.0 | 5.25 |
| HVOA, HV0B | -5.25 | -5.0 | -4.75 |
| LVOA, LV0B | 0 | 1.9 | V |
| HV2A, HV2B | -5 | -1.9 | 5 |
| LV2A, LV2B | 0 | 32 | 0 |
| HV1A, HV1B | -30 | -11.9 | 70 |
| LV1A, LV1B | $>H V 0$ and $>H V 2$ | 61 | V |
| VCWA, VCWB | -30 | -20.9 | $<$ LV0 and <LV2 |
| Maximum DELTA between HV1 to LV1 and HV2 to LV2 | 0 | 11 | V |
| INNx, INPx, EN, PDM, PCLKIN, U |  |  | 15 |
| INCWA, INCWB | 0 | V |  |
| INNxx, INPxx input sample rate | 0 | 5 | 100 |
| INNxx, INPXX input unit interval | 1 | V |  |
| PCLKIN input frequency | 10 | 10 | V |
| Ambient Temperature, TA | 1 | 100 | Msps |

## ELECTRICAL CHARACTERISTICS

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2=-11.9 \mathrm{~V}$, HV1 $=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted. The parameter results are applicable to both OUTA and OUTB, and they are measured using Non-Latch Mode unless otherwise noted.

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS | TEST LEVEL ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| HVO/LVO SIGNAL PATH - DC PERFORMANCE |  |  |  |  |  |  |
| P-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | HV0 $=2 \mathrm{~V}$, OUTX $=-750 \mathrm{mV}$ to -1.25 V | 6.5 | 9.5 | 13 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power, load $=100 \Omega$ to 0 V |  |  | 12\% |  | C |
| Output saturation current | $\mathrm{R}_{\mathrm{L}}=5 \Omega$ to -30 V | -3.1 | -1.3 | -1 | A | A |
| Output voltage |  |  | 1.0 |  | V | C |
| N-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | LV0 $=-2 \mathrm{~V}$, OUTX $=750 \mathrm{mV}$ to 1.25 V | 2.5 | 5 | 8.5 | $\Omega$ | A |
| Effective Resistance Variation | Max output power to Min output power, $\text { Load }=100 \Omega \text { to } 0 \mathrm{~V}$ |  |  | 5 | \% | C |
| Output saturation current | $\mathrm{R}_{\mathrm{L}}=5 \Omega$ to +30 V | 1.4 | 1.8 | 3.1 | A | A |
| Output voltage |  |  | -1.2 |  | V | C |
| HVO/LVO SIGNAL PATH - AC PERFORMANCE |  |  |  |  |  |  |
| Single-tone output frequency |  | 1 |  | 100 | Msps | B |
| $2^{\text {nd }}$ Order harmonic distortion (when using transformer bridge) | $\mathrm{f}=5.0 \mathrm{MHz}$ square wave, measured using transformer at secondary coil with $\mathrm{RL}=100 \Omega$ |  | 35 |  | dBc | C |
| $\mathrm{t}_{\mathrm{r}} \quad$ Output rise time | $10 \%$ to $90 \%$ of 0 V to + Vout Figure 8 |  | 4.5 |  | ns | C |
| $\mathrm{t}_{\mathrm{f}} \quad$ Output fall time | $10 \%$ to $90 \%$ of 0 V to -Vout Figure 8 |  | 1 |  | ns | C |
| $\mathrm{t}_{\mathrm{pr}}, \mathrm{t}_{\mathrm{pf}} \quad$ Propagation Delay | Input 50\% to Output 50\% Figure 8 |  | 30 |  | ns | B |
| HV2/LV2 SIGNAL PATH - DC PERFORMANCE |  |  |  |  |  |  |
| P-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | HV2 $=30 \mathrm{~V}$ to HV2 $=20 \mathrm{~V}$ | 4.5 | 9 | 12.5 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power, $\text { load }=100 \Omega \text { to } 0 \mathrm{~V}$ |  |  | 12\% |  | C |
| Output saturation current | HV2 $=60 \mathrm{~V} ; \mathrm{R}_{\mathrm{L}}=5 \Omega$ to GND | -4.1 | -2.3 | -1.8 | A | A |
| Output voltage |  |  | 28.5 |  | V | C |
| N-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | LV2 $=-10 \mathrm{~V}$ to $\mathrm{LV} 2=-12 \mathrm{~V}$ | 1.5 | 4.5 | 7.5 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power, $\text { load }=100 \Omega \text { to } 0 \mathrm{~V}$ |  |  | 4\% |  | C |
| Output saturation current | $\mathrm{LV} 2=-60 \mathrm{~V} ; \mathrm{R}_{\mathrm{L}}=5 \Omega$ to GND | 2.4 | 3.0 | 5.0 | A | A |
| Output Voltage |  |  | -10.5 |  | V | C |
| HV2/LV2 SIGNAL PATH - AC PERFORMANCE |  |  |  |  |  |  |
| Single-tone Output Frequency |  | 1 |  | 100 | Msps | B |
| $2^{\text {nd }}$ Order harmonic distortion when using transformer bridge | $\mathrm{f}=5.0 \mathrm{MHz}$ square wave, measured using transformer at secondary coil with $\mathrm{RL}=100 \Omega$ |  | 50 |  | dBc | C |
| $\mathrm{t}_{\mathrm{r}} \quad$ Output rise time | $10 \%$ to $90 \%$ of 0 V to + Vout Figure 8 |  | 7.5 |  | ns | C |
| $\mathrm{t}_{\mathrm{f}} \quad$ Output fall time | $10 \%$ to $90 \%$ of 0 V to -Vout Figure 8 |  | 3 |  | ns | C |
| $\mathrm{t}_{\mathrm{pr}}, \mathrm{t}_{\mathrm{pf}} \quad$ Propagation delay | Input 50\% to Output 50\% Figure 8 |  | 25 |  | ns | B |

(1) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.

## ELECTRICAL CHARACTERISTICS

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2$ $=-11.9 \mathrm{~V}, \mathrm{HV} 1=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted. The parameter results are applicable to both OUTA and OUTB, and they are measured using Non-Latch Mode unless otherwise noted.

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS | TEST LEVEL ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| HV1/LV1 SIGNAL PATH - DC PERFORMANCE |  |  |  |  |  |  |
| P-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | HV1 $=60 \mathrm{~V}$ to HV1 $=50 \mathrm{~V}$ | 2.5 | 7 | 12.5 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power load $=100 \Omega$ to GND |  |  | 11\% |  | C |
| Output saturation current | HV1 $=60 \mathrm{~V} ; \mathrm{R}_{\mathrm{L}}=5 \Omega$ to GND | -4.1 | -2.5 | -2 | A | A |
| Output voltage |  |  | 58 |  | V | C |
| N-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | LV1 $=-20 \mathrm{~V}$ to -10 V | 1 | 2 | 4.5 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power load $=100 \Omega$ to 0 V |  |  | 3\% |  | C |
| Output saturation current | $\mathrm{LV} 1=-60 \mathrm{~V} ; \mathrm{R}_{\mathrm{L}}=5 \Omega$ to GND | 2.9 | 3.4 | 4.1 | A | A |
| Output voltage |  |  | -20 |  | V | C |
| HV1/LV1 SIGNAL PATH - AC PERFORMANCE |  |  |  |  |  |  |
| Single-tone output frequency |  | 1 |  | 100 | Msps | B |
| $2^{\text {nd }}$ Order harmonic distortion (when using transformer bridge) | $\mathrm{f}=5.0 \mathrm{MHz}$ square wave, measured using transformer at secondary coil with $\mathrm{RL}=100 \Omega$ |  | 60 |  | dBc | C |
| $t_{r} \quad$ Output rise time | $10 \%$ to $90 \%$ of 0 V to + Vout Figure 8 |  | 6.5 |  | ns | C |
| $t_{f} \quad$ Output fall time | $10 \%$ to $90 \%$ of 0 V to - Vout Figure 8 |  | 3 |  | ns | C |
| $\mathrm{t}_{\mathrm{pr}}, \mathrm{t}_{\mathrm{pf}} \quad$ Propagation Delay | Input 50\% to Output 50\% Figure 8 |  | 25 |  | ns | B |

(1) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.

## ELECTRICAL CHARACTERISTICS

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2$ $=-11.9 \mathrm{~V}, \mathrm{HV} 1=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted. The parameter results are applicable to both OUTA and OUTB, and they are measured using Non-Latch Mode unless otherwise noted.

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS | TEST LEVEL ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CW SIGNAL PATH - DC PERFORMANCE |  |  |  |  |  |  |
| P-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | $\mathrm{VCW}=4.5 \mathrm{Vto} 5.5 \mathrm{~V}$ | 9 | 21 | 31 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power, $\text { load }=100 \Omega \text { to } 0 \mathrm{~V}$ |  |  | 30\% |  | C |
| Output saturation current | $\mathrm{R}_{\mathrm{L}}=5 \Omega$ to -20 V | -0.16 | -0.12 | -0.06 | A | A |
| Output voltage |  |  | 8 |  | V | C |
| N-CHANNEL |  |  |  |  |  |  |
| Effective resistance, RDSon + Rdiode | OUTX $=1 \mathrm{~V}$ to 2 V | 9 | 14 | 18 | $\Omega$ | A |
| Effective resistance variation | Max output power to Min output power, $\text { load }=100 \Omega \text { to } 0 \mathrm{~V}$ |  |  | 10\% |  | C |
| Output saturation current | $\mathrm{R}_{\mathrm{L}}=5 \Omega$ to 20 V | 0.29 | 0.35 | 0.44 | A | A |
| Output voltage |  |  | 30 |  | mV | C |

(1) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.

## ELECTRICAL CHARACTERISTICS (continued)

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2$ $=-11.9 \mathrm{~V}, \mathrm{HV} 1=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted. The parameter results are applicable to both OUTA and OUTB, and they are measured using Non-Latch Mode unless otherwise noted.

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS | TEST LEVEL ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CW SIGNAL PATH - AC PERFORMANCE ${ }^{(2)}$ |  |  |  |  |  |  |
| Single-tone output frequency |  | 0.5 |  | 10 | MHz | B |
| $2^{\text {nd }}$ Order harmonic distortion | $\mathrm{f}=1 \mathrm{MHz}$, measured using transformer at secondary coil with $R_{L}=100 \Omega$ |  | 47 |  | dBc | C |
|  | $\mathrm{f}=5 \mathrm{MHz}$, measured using transformer at secondary coil with $R_{L}=100 \Omega$ |  | 33 |  | dBc | C |
| Slew Rate + (Positive Edge) | $20 \%$ to $80 \%$ of Voutpp, measured using transformer at secondary coil with $\mathrm{RL}=100 \Omega$ |  | 0.6 |  | V/ns | c |
| Slew Rate - (Negative Edge) |  |  | 0.45 |  | V/ns | c |
| $\mathrm{tr}_{\text {r }} \quad$ Output rise time | $10 \%$ to $90 \%$ of 0 V to + Vout Figure 8 |  | 30 |  | ns | C |
| $\mathrm{t}_{\mathrm{f}} \quad$ Output fall time | $10 \%$ to $90 \%$ of 0 V to - Vout Figure 8 |  | 10 |  | ns | C |
| $\mathrm{t}_{\mathrm{p},}, \mathrm{t}_{\mathrm{pf}}$ Propagation Delay | Input 50\% to Output 50\% Figure 8 |  | 25 |  | ns | B |
| AC-coupled gate drive time constant for P-CHANNEL |  | 10 | 20 | 30 | $\mu \mathrm{s}$ | C |
| CW InPUT CHARACTERISTIC |  |  |  |  |  |  |
| High input voltage |  | 1.05 |  |  | V | B |
| Low input voltage |  |  |  | 0.35 | V | B |
| Low input current | CWINX=0V |  | 0 | 1 | $\mu \mathrm{A}$ | B |
| High input current | CWINX=5.0V |  | 25 | 40 | $\mu \mathrm{A}$ | B |
| Input Gate Charge | $\begin{aligned} & \text { CWINX }=0 \mathrm{~V} \text { to } 5.0 \mathrm{~V} \text { or } \\ & 5.0 \mathrm{~V} \text { to } 0 \mathrm{~V} \end{aligned}$ |  | 550 |  | pC | C |
| {LOGIC CHARACTERISTICS - INNXX, INPXX, EN |  |  |  |  |  |  |
| , PDM ${ }^{\text {, PCLKIN pins }}$} |  |  |  |  |  |  |
| Input capacitance | INNxx, INPxx, PCLKIN @ 10 MHz |  | 6 |  | pF | C |
|  | EN $@ 10 \mathrm{MHz}$ |  | 9 |  |  |  |
|  | PDM @ 10 MHz |  | 4 |  |  |  |
| Logic high input voltage | $\mathrm{VAA}=2.375 \mathrm{~V}$ to 3.6 V | $0.55 *$ VAA |  | VAA | V | B |
| Logic low input voltage | $\mathrm{VAA}=2.375 \mathrm{~V}$ to 3.6 V | 0 |  | 0.8 | V | B |
| Logic low input current |  |  | 0.2 | 10 | $\mu \mathrm{A}$ | B |
| Logic high input current |  |  | 0.2 | 10 | $\mu \mathrm{A}$ | B |
| Minimum clock period, tper | Figure 9, PCLKIN | 10 |  |  | ns | B |
| Minimum clock high time, tmin | Figure 9, PCLKIN | 2.0 |  |  | ns | B |
| $\mathrm{t}_{\text {s }} \quad$ Setup time | Figure 9, PCLKIN, INNxx, INPxx | 0 |  |  | ns | B |
| $\mathrm{t}_{\mathrm{h}} \quad$ Hold time | Figure 9, PCLKIN, INNxx, INPxx | 1.5 |  |  | ns | B |
| OUTPUT CHARACTERISTIC |  |  |  |  |  |  |
| Output resistance | Power Down Mode (Hi-Z Output) VTEST $=20 \mathrm{~V}$ |  | 1 |  | G $\Omega$ | c |
| Output capacitance | Power Down Mode (Hi-Z Output) @1 to 100 MHZ |  | 165 |  | pF | C |
| Leakage current | Power Down Mode (Hi-Z Output) VTEST $=0 \mathrm{~V}$ |  | 0.001 | 10 | $\mu \mathrm{A}$ | A |
| INTERNAL GATE CHARGE CHARACTERISTICS |  |  |  |  |  |  |
| Input gate charge ${ }^{(3)}$ | HV0/LV0 internal FET gates driven from VEE to VDD or VDD to VEE |  | 3.5 |  | nC | C |
|  | HV1/LV1 internal FET gates driven from VEE to VDD or VDD to VEE |  | 4.6 |  | nC | C |
|  | HV2/LV2 internal FET gates driven rom VEE to VDD or VDD to VEE |  | 7 |  | nC | C |

(1) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.
(2) TX517 CW outputs are complimentary. Thus a transformer is needed to enable CW output.
(3) Input gate charge is the amount of charge to change the internal FET gates of a given output from either a low to a high state or from a high to a low state. Each gate charge value applies to both the $P$ and $N$ type FET for the given output. These values can be used to estimate the amount of dynamic current that needs to be provided to the VDD and VEE power supplies in order to switch the internal FET's at a given sampling rate.

## ELECTRICAL CHARACTERISTICS

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2$ $=-11.9 \mathrm{~V}, \mathrm{HV} 1=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted. The parameter results are applicable to both OUTA and OUTB, and they are measured using Non-Latch Mode unless otherwise noted.

| PARAMETER | CONDITIONS |  | MIN | TYP | MAX | UNITS | TEST <br> LEVEL ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| POWER SUPPLY |  |  |  |  |  |  |  |
| Total Quiescent Current (PW Mode) Power supply VDD | $\operatorname{INPxx}=1, \operatorname{INNxx}=0$, PCLKIN $=0$ or 1 |  |  | 13 | 15 | mA | A |
| Total Quiescent Current (PW Mode) Power supply VEE | $\operatorname{INPxx}=1, \operatorname{INNxx}=0, \mathrm{PCLKIN}=0$ or 1 |  | -10 | -8 |  | mA | A |
| Total Quiescent Current (PW Mode) Power supply VAA | INPxx = 1, INNxx = 0, PCLKIN=0 or 1 |  | -3 | -2 |  | mA | A |
| Dynamic Current Consumption (PW Mode) Power supply VDD | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at $10 \mathrm{Msps}(5 \mathrm{MHz})$ on noted signal path. Load $=$ transformer and 100 ohm differential load, see Figure 10. | HV0/LV0 |  | 17 | 23 | mA | B |
|  |  | HV1/LV1 |  | 18 | 23 |  |  |
|  |  | HV2/LV2 |  | 20.5 | 23 |  |  |
| Dynamic Current Consumption (PW Mode) Power supply VEE | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at $10 \mathrm{Msps}(5 \mathrm{MHz})$ on noted signal path. Load $=$ transformer and 100 ohm differential load, see Figure 10. | HV0/LV0 | -15 | -10 |  | mA | B |
|  |  | HV1/LV1 | -15 | -10.5 |  |  |  |
|  |  | HV2/LV2 | -15 | -12.5 |  |  |  |
| Dynamic Current Consumption (PW Mode) Power supply VAA | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at $10 \mathrm{Msps}(5 \mathrm{MHz})$ on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. | HV0/LV0 | -4 | -2.3 |  | mA | B |
|  |  | HV1/LV1 | -4 | -2.5 |  |  |  |
|  |  | HV2/LV2 | -4 | -2.5 |  |  |  |
| Dynamic Current Consumption (PW Mode) Power supply HV0 | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 Msps ( 5 MHz ) on noted signal path. Load $=$ transformer and 100 ohm differential load, see Figure 10. |  |  | 2 | 4 | mA | B |
| Dynamic Current Consumption (PW Mode) Power supply LV0 | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 Msps ( 5 MHz ) on noted signal path. Load $=$ transformer and 100 ohm differential load, see Figure 10. |  | -3.5 | -2 |  | mA | B |
| Dynamic Current Consumption (PW Mode) Power supply HV1 | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 Msps ( 5 MHz ) on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. |  |  | 41 | 60 | mA | B |
| Dynamic Current Consumption (PW Mode) Power supply LV1 | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 Msps ( 5 MHz ) on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. |  | -55 | -41 |  | mA | B |
| Dynamic Current Consumption (PW Mode) Power supply HV2 | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 $\operatorname{Msps}(5 \mathrm{MHz})$ on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. |  |  | 22 | 60 | mA | B |
| Dynamic Current Consumption (PW Mode) Power supply LV2 | Input pattern = 10 cycle square wave, $5 \%$ ( 5 MHz ) on noted signal path. Load $=$ trans differential load, see Figure 10. | duty cycle at 10 Msps former and 100 ohm | -35 | -22 |  | mA | B |
| Total Power Dissipation for device only (PW Mode) | Input pattern = 10 cycle square wave, $5 \%$ duty cycle at 10 Msps on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. | HV0/LV0 |  | 0.15 | 0.25 | W | B |
|  |  | HV1/LV1 |  | 1.1 | 1.7 |  |  |
|  |  | HV2/LV2 |  | 0.6 | 0.8 |  |  |
| Dynamic Current Consumption (CW Mode) <br> Power supply VCWA + VCWB | Input pattern $=10$ cycle square wave, $100 \%$ duty cycle at 10 Msps on CW signal path. Load = transformer and 100 ohm differential load, see Figure 10. $\mathrm{EN} \backslash=0$ or $1, \mathrm{PCLKIN}=0$ or 1 |  |  | 62 | 100 | mA | B |
| Total Power Dissipation for device only (CW Mode) | Input pattern $=10$ cycle square wave, $100 \%$ duty cycle at 10 Msps ( 5 MHz ) on noted signal path. Load = transformer and 100 ohm differential load, see Figure 10. $\mathrm{EN}=0$ or $1, \mathrm{PCLKIN}=0$ or 1 |  |  | 310 | 400 | mW | B |
| Supply (HVx, LVx) Slew Rate Limit |  |  |  |  | 10 | V/ms | B |
| POWER-DOWN CHARACTERISTIC |  |  |  |  |  |  |  |
| Power-Down Dissipation | Power Down Mode (Hi-Z Output) PDM $=0, I N P x x=1, I N N x x=0$ PCLKIN $=0$ or 1 |  |  | 3 | 15 | mW | A |

[^0]
## ELECTRICAL CHARACTERISTICS (any level to any level transitions - 17 level output, 289 unique transitions ${ }^{(1)}$ )

All Specifications at: $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{VAA}=2.5 \mathrm{~V}, \mathrm{VDD}=5 \mathrm{~V}, \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2$ $=-11.9 \mathrm{~V}, \mathrm{HV} 1=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted.

| PARAMETER | CONDITIONS | MIN TYP | MAX | UNITS | $\begin{gathered} \text { TEST } \\ \text { LEVEL }^{(2)} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| POWER UP/DOWN TIMING |  |  |  |  |  |
| Power down time |  | 100 |  | ns | C |
| Power up time |  | 100 |  | ns | C |
| HVX/LVX SIGNAL PATH - AC PERFORMANCE |  |  |  |  |  |
| Mean normalized output rise time | $10 \%$ to $90 \%$ of 0 to $1,20 \mathrm{MHz}$ | 5 |  | ns | C |
| Mean delay (relative to clock edge of $1^{\text {st }}$ sample) | $0-20 \mathrm{MHz}$ | 23 |  | ns | C |
| Delay standard deviation | $0-20 \mathrm{MHz}$ | 1.2 |  | ns | C |
| Phase standard deviation | 5 MHz | 0.01 |  | cycles | C |
|  | 20 MHz | 0.03 |  | cycles | C |
| Gain standard deviation | 5 MHz | 4 |  | \% | C |
|  | 20 MHz | 8 |  | \% | C |

(1) These parameters are measured on the differential output starting from 1 of 17 possible states to every other possible state. Therefore, $17 \mathrm{X17}=289$ unique transitions.
(2) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.

## TYPICAL CHARACTERISTICS

All Specifications at: $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{VAA}=+2.5 \mathrm{~V}, \mathrm{VDD}=+5 \mathrm{~V} . \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV}=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2=-11.9 \mathrm{~V}$, HV1 $=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega$ to $G N D$ for OUTA, $\mathrm{R}_{\mathrm{L}}=100 \Omega$ to GND for OUTB, unless otherwise noted.


Figure 1. 17-level Outputs with 10ns Pulse Width (100MSPS)


Figure 2. 17-level Outputs with 20ns Pulse Width (50MSPS)


Figure 3. 5MHz 3-level 10 Cycles Outputs

## TYPICAL CHARACTERISTICS (continued)

All Specifications at: $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{VAA}=+2.5 \mathrm{~V}, \mathrm{VDD}=+5 \mathrm{~V} . \mathrm{VEE}=-5 \mathrm{~V}, \mathrm{HV} 0=1.9 \mathrm{~V}, \mathrm{LV} 0=-1.9 \mathrm{~V}, \mathrm{HV} 2=32 \mathrm{~V}, \mathrm{LV} 2=-11.9 \mathrm{~V}$, HV1 $=+61.1 \mathrm{~V}, \mathrm{LV} 1=-20.9 \mathrm{~V}, \mathrm{VCW}=11 \mathrm{~V}, R_{L}=100 \Omega$ to GND for OUTA, $R_{L}=100 \Omega$ to GND for OUTB, unless otherwise noted.


Figure 4. 3-level Outputs with 100ns Pulse Width (10MSPS)


Figure 5. 5MHz 5-level Outputs


Figure 6. 2 MHz CW Outputs

## APPLICATION INFORMATION

Table 1. Truth Table

| Description | $\overline{\text { EN }}$ | $\overline{\text { PDM }}$ | PCLKIN | CWINA | CWINB | INPxx ${ }^{(1)}$ | INNxx ${ }^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power Down (Hi-Z Output) | 1 | 0 | $\mathrm{x}^{(2)}$ | 0 | 0 | 1 | 0 |
| CW Mode | x | 0 | x | $0 / 1$ | $1 / 0$ | 1 | 0 |
| Non-Latch Mode | 1 | 1 | x | 0 | 0 | $0 / 1$ | $0 / 1$ |
| Latch Mode | 0 | 1 | $0 / 1$ | 0 | 0 | $0 / 1$ | $0 / 1$ |

(1) The logic device driving the inputs of the TX517 should include means to prevent a 'shoot-thru' fault condition. Any input combination that would result in an INP-input to be Low (0) and an INN-input to be High (1) at the same time on the same output (OUTA or OUTB) could result in permanent damage to the TX517. See also disallowed logic state table. Table 3 is provided for an example of how to properly drive the TX517 data inputs INPxx and INNxx.
(2) $X=$ don't care state. However, in order to prevent excessive power consumption it is recommended that all unused inputs be tied off to a logic high or logic low. The logic inputs to the device have no internal tie-off's.

Table 2. Disallowed Logic States

| Description | $\overline{\text { EN }}$ | $\overline{\text { PDM }}$ | PCLKIN | CWINA | CWINB | INPxA | INN $\mathbf{x A}$ | INPxB | INNxB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Disallowed mode 1 ${ }^{(1)}$ | x | x | x | x | x | 0 | 1 | x | x |
| Disallowed mode 2 |  |  |  |  |  |  |  |  |  |
| (1) | x | x | x | x | x | x | x | 0 | 1 |
| Disallowed mode 3 ${ }^{(2)}$ | x | 0 | x | x | x | x | 1 | x | x |
| Disallowed mode 4 ${ }^{(2)}$ | x | 0 | x | x | x | x | x | x | 1 |
| Disallowed mode 5 ${ }^{(2)}$ | x | 0 | x | x | x | 0 | x | x | x |
| Disallowed mode 6 ${ }^{(2)}$ | x | 0 | x | x | x | x | x | 0 | x |
| Disallowed mode $7^{(3)}$ | 0 | x | 0 | x | x | x | x | x | x |

(1) This logic state causes a 'shoot-thru' fault condition that could result in permanent damage to the TX517.
(2) This logic state causes a high power consumption condition in the internal logic circuitry of the TX517 and could result in a long term reliability failure of the TX517.
(3) This disallowed logic state is only valid for DC conditions. i.e. it is not allowed to keep PCLKIN at a low logic state when EN is at a low logic state. This causes a high power consumption condition in the internal logic circuitry of the TX517. However, it is acceptable to drive $\mathrm{EN} \backslash$ low and drive PCLKIN with a clock waveform under the recommended operating conditions for PCLKIN.

Table 3. Example Input Data Set of a 17-Level Output ${ }^{(1)}$

| Output Level | INPOA | INP2A | INP1A | INP1B | INP2B | INPOB | INNOA | INN2A | INN1A | INN1B | INN2B | INNOB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 7 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| 6 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 |
| 5 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 4 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| 3 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 |
| 2 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 |
| 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 1 |
| -1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 1 |
| -2 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 |
| -3 | 1 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| -4 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| -5 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| -6 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| -7 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| -8 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| off state | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(1) The levels listed in this table are active high; the $P$ signals need to be inverted before driving the chip. This note is only applicable to THIS particular table ("the example input data set of a 17-level output).

Table 4. Power Supplies Sequence

|  | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :--- | :--- | :--- | :--- | :--- |
| Power-up | Driver Supplies(VEE, VAA, VDD) | LV1 | HV1 | LV2, LV0, HV0, HV2, VCW |
| Power-down | VCW, HV2, HV0, LV0, LV2 | HV1 | LV1 | Driver Supplies (VDD, VAA, VEE) |


A. Diodes D1, D2 placeholders only; choose appropriate model (e.g. MMBD3004S)
B. Load resistor R1, and capacitor C11 usage and values may vary depending on final configuration
C. Bypass capacitors and values on all supplies are placeholders only. Capacitors between various supply rails may also be necessary.
D. Inductors (ferrite beads) L1, L2 are optional components
E. Voltages levels on the voltage supplies correspond to the ones used at simulation

## Figure 7. Typical Device Configuration

BLOCK DIAGRAM


TIMING RELATED INFORMATION


Figure 8. Output Timing Information


Figure 9. Timing Waveform for Latch Mode


Note A: This signal is supplied by a function generator with the following characteristics: 0 to 2.5 V square wave, $\mathrm{tr} / \mathrm{tf}<3 n s$, frequency as noted in the electrical characteristics.

Figure 10. Loading for Power Consumption Tests

## REVISION HISTORY

## Changes from Original (September 2011) to Revision A

- Fixed duty cycle typo, changed duty cycle from $5 \%$ to $100 \%$ for "Dynamic Current Consumption (CW Mode) Power supply VCWA + VCWB" in the ELECTRICAL CHARACTERISTICS table.
- Fixed duty cycle typo, changed duty cycle from $5 \%$ to $100 \%$ for "Total Power Dissipation for device only (CW Mode)" in the ELECTRICAL CHARACTERISTICS table.


## PACKAGING INFORMATION

| Orderable Device | Status <br> (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan <br> (2) | Lead finish/ Ball material <br> (6) | MSL Peak Temp <br> (3) | Op Temp ( ${ }^{\circ} \mathrm{C}$ ) | Device Marking <br> (4/5) | Samples |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TX517IZCQ | ACTIVE | NFBGA | ZCQ | 144 | 160 | RoHS \& Green | Call TI | Level-3-260C-168 HR | 0 to 85 | TX5171 | Samples |

${ }^{(1)}$ The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.
${ }^{(2)}$ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed $0.1 \%$ by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: Tl defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement
${ }^{(3)}$ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
${ }^{(4)}$ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
${ }^{(5)}$ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
${ }^{(6)}$ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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## TRAY



Chamfer on Tray corner indicates Pin 1 orientation of packed units.
*All dimensions are nominal

| Device | Package <br> Name | Package <br> Type | Pins | SPQ | Unit array <br> matrix | Max <br> temperature <br> $\left({ }^{\circ} \mathbf{C}\right)$ | L (mm) | W <br> $(\mathbf{m m})$ | K0 <br> $(\boldsymbol{\mu m})$ | P1 <br> $(\mathbf{m m})$ | $\mathbf{C L}$ <br> $(\mathbf{m m})$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TX517IZCQ | ZCQ | NFBGA | 144 | 160 | $8 \times 20$ | 150 | 315 | 135.9 | 7620 | 15 | 15 |
| $(\mathbf{m m})$ |  |  |  |  |  |  |  |  |  |  |  |



NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.


NOTES: (continued)
3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints For information, see Texas Instruments literature number SPRAA99 (www.ti.com/lit/spraa99).


SOLDER PASTE EXAMPLE BASED ON 0.15 mm THICK STENCIL SCALE:6X

NOTES: (continued)
4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

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[^0]:    (1) Test levels: (A) $100 \%$ tested at $25^{\circ} \mathrm{C}$. Over temperature limits by characterization and simulation. (B) Limits set by characterization and simulation. (C) Typical value only for information.

